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**Technical Report**  
RAL-TR-1998-086

# Some Noise Calculations for Time Invariant Filters

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18<sup>th</sup> January 1999

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**ISSN 1358-6254**

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**Some Noise Calculations for time invariant filters.**  
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This paper presents the noise calculations for some detector readout circuits following the method outlined in Ref 1 and Ref 2. The results are summarised in Ref 3. The circuits considered are a charge amplifier combined with the following filters :-

RC  
 RC-CR  
 CR-RC<sup>2</sup>  
 CR-RC<sup>n</sup>  
 CR<sup>2</sup>-RC (bipolar)  
 Indefinite Cusp  
 Trapezoidal  
 CR followed by Correlated Double Sampling  
 Triple Sampled Deconvolution.

The voltage-to-voltage transfer functions for the filters are shown in Appendix A.. The calculations of the thermal, shot and flicker noise for each filter are given in Appendix B.. The calculations are performed in the time domain and then repeated in the frequency domain. The calculation of flicker noise is not done in the time domain. The current-to-voltage transfer function and impulse response from the noise current source in the front end device of the charge amplifier to the output, are given by

$$H_r(s) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} H_v(s)$$

$$h_r(t_{m1} - t_p) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} h_v(t_{m1} - t_p)$$

respectively. The voltage-to-voltage transfer function of the filters are given by  $H_v(s)$ . The voltage-to-voltage impulse response of the filters are given by  $h_v(t_{m1}-t_p)$ . The current-to-voltage transfer function and impulse response from the noise current source of the detector leakage current to the output, are given by

$$H_i(s) = \frac{A}{(c_{tot} + A c_f)} G_v(s)$$

$$h_i(t_{m1} - t_p) = \frac{A}{(c_{tot} + A c_f)} g_v(t_{m1} - t_p)$$

respectively. The voltage-to-voltage step response function of the filters are given by  $G_v(s)$ . The voltage-to-voltage step response of the filters are given by  $g_v(t_{m1}-t_p)$ .

**References.**

- [1] P. Seller, Noise analysis in linear electronic circuits. Nucl. Inst. and Meth. A 376 (1996) 229-241.
- [2] P. Seller, Erratum to "Noise analysis in linear electronic circuits". Nucl. Inst. and Meth. A 408 (1998) 603-604.
- [3] P. Seller, Summary of thermal, shot and flicker noise in detectors and readout circuits. Nucl. Inst. and Meth. (1996) To be published.

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A. The voltage-to-voltage responses.

$$RC \quad H_v(s) = \frac{1}{1+s\tau}$$

$$h_v(t_{m1}-t_p) = \frac{1}{\tau} e^{-\frac{(t_{m1}-t_p)}{\tau}}$$

$$G_v(s) = \frac{1}{s(1+s\tau)}$$

$$g_v(t_{m1}-t_p) = 1 - e^{-\frac{(t_{m1}-t_p)}{\tau}} \quad \text{peak value} = 1 \text{ at } (t_{m1}-t_p) = \infty$$

$$CR \quad H_v(s) = \frac{s\tau}{1+s\tau}$$

$$h_v(t_{m1}-t_p) = \delta(0) - \frac{1}{\tau} e^{-\frac{(t_{m1}-t_p)}{\tau}}$$

$$G_v(s) = \frac{\tau}{1+s\tau}$$

$$g_v(t_{m1}-t_p) = e^{-\frac{(t_{m1}-t_p)}{\tau}} \quad \text{peak value} = 1 \text{ at } (t_{m1}-t_p) = 0$$

$$CR-RC \quad H_v(s) = \frac{s\tau}{(1+s\tau)^2}$$

$$h_v(t_{m1}-t_p) = \left( \frac{1}{\tau} - \frac{(t_{m1}-t_p)}{(\tau)^2} \right) e^{-\frac{(t_{m1}-t_p)}{\tau}}$$

$$G_v(s) = \frac{\tau}{(1+s\tau)^2}$$

$$g_v(t_{m1}-t_p) = \frac{(t_{m1}-t_p)}{\tau} e^{-\frac{(t_{m1}-t_p)}{\tau}} \quad \text{peak value} = \frac{1}{e} \text{ at } (t_{m1}-t_p) = \tau$$

$$CR-RC^2 \quad H_v(s) = \frac{s\tau}{(1+s\tau)^3}$$

$$h_v(t) = \left( \frac{t_{m1}-t_p}{(\tau)^2} - \frac{(t_{m1}-t_p)^2}{2(\tau)^3} \right) e^{-\frac{(t_{m1}-t_p)}{\tau}}$$

$$G_v(s) = \frac{\tau}{(1+s\tau)^3}$$

$$g_v(t) = \frac{(t_{m1}-t_p)^2}{2(\tau)^2} e^{-\frac{(t_{m1}-t_p)}{\tau}} \quad \text{peak value} = \frac{2}{e^2} \text{ at } (t_{m1}-t_p) = 2\tau$$

A. The voltage-to-voltage responses.

$$CR - RC^n \quad H_v(s) = \frac{s\tau}{(1+s\tau)^{n+1}}$$

$$h_v(t) = \left( \frac{(t_{m1} - t_p)^{n-1}}{(\tau)^n (n-1)!} - \frac{(t_{m1} - t_p)^n}{(\tau)^{n+1} n!} \right) e^{-\frac{(t_{m1} - t_p)}{\tau}}$$

$$G_v(s) = \frac{\tau}{(1+s\tau)^{n+1}}$$

$$g_v(t_{m1} - t_p) = \frac{(t_{m1} - t_p)^n}{(\tau)^n n!} e^{-\frac{(t_{m1} - t_p)}{\tau}}$$

$$\text{peak value} = \frac{n^n}{n! e^n} \text{ at } (t_{m1} - t_p) = n\tau$$

$$CR^2 - RC \quad H_v(s) = \frac{(s\tau)^2}{(1+s\tau)^3}$$

$$h_v(t_{m1} - t_p) = \left( \frac{1}{\tau} - \frac{2(t_{m1} - t_p)}{(\tau)^2} + \frac{(t_{m1} - t_p)^2}{2(\tau)^3} \right) e^{-\frac{(t_{m1} - t_p)}{\tau}}$$

$$G_v(s) = \frac{s(\tau)^2}{(1+s\tau)^3}$$

$$g_v(t_{m1} - t_p) = \left( \frac{(t_{m1} - t_p)}{\tau} - \frac{(t_{m1} - t_p)^2}{2(\tau)^2} \right) e^{-\frac{(t_{m1} - t_p)}{\tau}}$$

$$\text{peak at } (t_{m1} - t_p) = (2 \pm \sqrt{2})\tau \quad \text{crosses zero at } (t_{m1} - t_p) = 2\tau$$

$$\text{positive peak} = (-1 + \sqrt{2})e^{-2+\sqrt{2}} = 0.23 = \frac{1}{4.3} \quad \text{negative peak} = -0.079$$

A. The voltage-to-voltage responses.

Cusp 
$$H_v(j\omega) = \frac{2\omega\tau}{1 + (\omega\tau)^2}$$

For time invariant filter peaking at  $t_{m1} - t_s$  ( $t_s$  will tend to  $\infty$  later in noise calculations)

$$h_v(t_{m1} - t_p) = \frac{1}{\tau} \left( 1 - e^{-\frac{t_s}{\tau}} \right) e^{-\frac{(t_{m1} - (t_p + t_s))}{\tau}} \quad t_0 \langle t_p \rangle t_{m1} - t_s$$

$$h_v(t_{m1} - t_p) = -\frac{1}{\tau} \left( e^{\frac{(t_p + t_s - t_{m1})}{\tau}} - e^{-\frac{t_s}{\tau}} \right) \quad t_{m1} - t_s \langle t_p \rangle t_{m1}$$

$$G_v(j\omega) = \frac{2\tau}{1 + (\omega\tau)^2}$$

$$g_v(t_{m1} - t_p) = \left( 1 - e^{-\frac{t_s}{\tau}} \right) e^{-\frac{(t_{m1} - (t_p + t_s))}{\tau}} \quad t_0 \langle t_p \rangle t_{m1} - t_s$$

$$g_v(t_{m1} - t_p) = \left( e^{\frac{(t_p + t_s - t_{m1})}{\tau}} - e^{-\frac{t_s}{\tau}} \right) \quad t_{m1} - t_s \langle t_p \rangle t_{m1}$$

peak value =  $(1 - e^{-t_s})$  at  $t_p = t_{m1} - t_s$

Trapezoidal 
$$|H(j\omega)| = \frac{4}{\omega t_{rise}} \sin \frac{\omega t_{rise}}{2} \sin \left( \frac{t_{rise} + t_{flat}}{2} \right) \quad (\text{if } t_{rise} = t_{fall})$$

$$h_v(t_{m1} - t_p) = 0 \quad t_0 \langle t_p \rangle t_{m1} - (t_{rise} + t_{flat} + t_{fall})$$

$$h_v(t_{m1} - t_p) = \frac{1}{t_{fall}} \quad t_{m1} - (t_{rise} + t_{flat} + t_{fall}) \langle t_p \rangle t_{m1} - (t_{rise} + t_{flat})$$

$$h_v(t_{m1} - t_p) = 0 \quad t_{m1} - (t_{rise} + t_{flat}) \langle t_p \rangle t_{m1} - t_{rise}$$

$$h_v(t_{m1} - t_p) = \frac{-1}{t_{rise}} \quad t_{m1} - t_{rise} \langle t_p \rangle t_{m1}$$

$$|G(j\omega)| = \frac{4}{\omega^2 t_{rise}} \sin \frac{\omega t_{rise}}{2} \sin \left( \frac{t_{rise} + t_{flat}}{2} \right) \quad (\text{if } t_{rise} = t_{fall})$$

$$g_v(t_{m1} - t_p) = 0 \quad t_0 \langle t_p \rangle t_{m1} - (t_{rise} + t_{flat} + t_{fall})$$

$$g_v(t_{m1} - t_p) = \frac{(t_p + t_{rise} + t_{flat} + t_{fall}) - t_{m1}}{t_{fall}} \quad t_{m1} - (t_{rise} + t_{flat} + t_{fall}) \langle t_p \rangle t_{m1} - (t_{rise} + t_{flat})$$

$$g_v(t_{m1} - t_p) = 1 \quad t_{m1} - (t_{rise} + t_{flat}) \langle t_p \rangle t_{m1} - t_{rise}$$

$$g_v(t_{m1} - t_p) = \frac{t_{m1} - t_p}{t_{rise}} \quad t_{m1} - t_{rise} \langle t_p \rangle t_{m1}$$

peak value = 1

A. The voltage-to-voltage responses.

RC followed by Correlated Double Sampler (samples spaced at  $\Delta = t_{m2} - t_{m1}$ )

$$b_1 = -1, b_2 = 1$$

$$H_v(s) = \frac{1}{1 + s\tau} (1 - e^{-s\Delta}) \quad |H_v(j\omega)|^2 = \frac{4 \sin^2\left(\frac{\omega\Delta}{2}\right)}{(1 + \omega^2\tau^2)} = \frac{2(1 - \cos \omega\Delta)}{(1 + \omega^2\tau^2)}$$

$$G_v(s) = \frac{1}{s(1 + s\tau)} (1 - e^{-s\Delta}) \quad |H_v(j\omega)|^2 = \frac{4 \sin^2\left(\frac{\omega\Delta}{2}\right)}{\omega^2(1 + \omega^2\tau^2)} = \frac{2(1 - \cos \omega\Delta)}{\omega^2(1 + \omega^2\tau^2)}$$

if input signal is at  $t_{m1}$  then the output grows as  $\left(1 - e^{-\frac{t_{m1}}{\tau}}\right)$  and peak value = 1

RC - CR - Triple sample deconvolution. Samples spaced at  $\Delta$  and second sample on start of signal.

$$b_1 = \frac{\tau}{\Delta} e^{-\frac{\tau+\Delta}{\tau}}, b_2 = -\frac{2\tau}{\Delta} e^{-\frac{\tau}{\tau}}, b_3 = \frac{\tau}{\Delta} e^{-\frac{\tau-\Delta}{\tau}}$$

$$|H_v(j\omega)|^2 = \frac{\omega^2\tau^2}{(1 + \omega^2\tau^2)^2} \left(\frac{\tau}{e\Delta}\right)^2 \left(e^{\left(\frac{\Delta}{\tau}\right)} + e^{-\left(\frac{\Delta}{\tau}\right)} - 2 \cos \omega\Delta\right)^2$$

$$\text{signal gain} = 0b_1 + 0b_2 + \frac{\Delta}{\tau} e^{-\left(\frac{\Delta}{\tau}\right)} b_3 = e^{-1}$$

Zero ohm resistor

$$H_v(s) = 1 \quad h_v(t_{m1} - t_p) = \delta(t_{m1} - t_p)$$

$$G_v(s) = \frac{1}{s} \quad g_v(t_{m1} - t_p) = u(t_{m1} - t_p)$$

B. Thermal noise through a charge amp and RC filter in time domain.

$$\sigma_{V_{out}}^2(t_{01}) = \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_i h_r^2(t_m - t_p) dt_p$$

$$h_r(t_m - t_p) = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \frac{1}{\tau} e^{-\frac{t_m - t_p}{\tau}}$$

$$\sigma_{V_{out}}^2(t_{01}) = \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{\tau^2} e^{-\frac{2(t_{m1} - t_p)}{\tau}} dt_p$$

$$\sigma_{V_{out}}^2(t_{01}) = \pi Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{\tau^2} \frac{\tau}{2} \left[ e^{-\frac{2(t_{m1} - t_p)}{\tau}} \right]_{t_0}^{t_{m1}}$$

$$\sigma_{V_{out}}^2(t_{01}) = \frac{\pi Q_i}{2\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( 1 - e^{-\frac{2(t_{m1} - t_0)}{\tau}} \right)$$

$$\sigma_{V_{out}}^2(t_{\infty}) = \frac{kTr}{\tau} \left( \frac{Ac_{tot}}{c_{tot} + Ac_f} \right)^2$$



B. Thermal noise through a charge amp and RC filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_{\infty}) = \int_0^{\infty} Q_t |H_r(jw)|^2 dw$$

$$H_r(s) = \frac{Ac_{tot}}{gm(c_{tot} + Ac_f)} \frac{1}{(1 + s\tau)}$$

$$\sigma_{V_{out}}^2(t_{\infty}) = \int_0^{\infty} Q_t \left( \frac{Ac_{tot}}{gm(c_{tot} + Ac_f)} \right)^2 \frac{1}{1 + (w\tau)^2} dw$$

$$\sigma_{V_{out}}^2(t_{\infty}) = Q_t \left( \frac{Ac_{tot}}{gm(c_{tot} + Ac_f)} \right)^2 \frac{1}{\tau} [\tan^{-1}(w\tau)]_0^{\infty}$$

$$\sigma_{V_{out}}^2(t_{\infty}) = \left( \frac{Ac_{tot}}{gm(c_{tot} + Ac_f)} \right)^2 \frac{Q_t \pi}{\tau 2}$$

$$\sigma_{V_{out}}^2(t_{\infty}) = \left( \frac{Ac_{tot}}{(c_{tot} + Ac_f)} \right)^2 \frac{kTr}{\tau}$$

B. Shot noise through a charge amp and RC filter in the time domain.

$$\sigma_{Vout}^2(t_{01}) = \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_s h_i^2(t_m - t_p) dt$$

$$h_i(t_m - t_p) = \frac{A}{c_{tot} + A c_f} \left( 1 - e^{-\frac{t_m - t_p}{\tau}} \right)$$

$$\sigma_{Vout}^2(t_{01}) = \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( 1 - e^{-\frac{t_{m1} - t_p}{\tau}} \right)^2 dt_p$$

$$\sigma_{Vout}^2(t_{01}) = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \int_{t_p=t_0}^{t_p=t_{m1}} \left( 1 - 2e^{-\frac{(t_{m1} - t_p)}{\tau}} + e^{-\frac{2(t_{m1} - t_p)}{\tau}} \right) dt_p$$

$$\sigma_{Vout}^2(t_{01}) = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left[ t_p - 2\tau e^{-\frac{(t_{m1} - t_p)}{\tau}} + \frac{\tau}{2} e^{-\frac{2(t_{m1} - t_p)}{\tau}} \right]_{t_0}^{t_{m1}}$$

$$\sigma_{Vout}^2(t_{01}) = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( t_{m1} - 2\tau + \frac{\tau}{2} - t_0 + 2\tau e^{-\frac{t_{m1} - t_0}{\tau}} - \frac{\tau}{2} e^{-\frac{2(t_{m1} - t_0)}{\tau}} \right)$$

calling  $t_{01} = t_{m1} - t_0$

$$\sigma_{Vout}^2(t_{01}) = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( t_{01} - \frac{3\tau}{2} + 2\tau e^{-\frac{t_{01}}{\tau}} - \frac{\tau}{2} e^{-\frac{2(t_{01})}{\tau}} \right)$$

If  $t_{01}$  is large then this tends to

$$\sigma_{Vout}^2(t_{01}) = i q \left( \frac{A}{c_{tot} + A c_f} \right)^2 (t_{01})$$

B. Flicker noise through a charge amp and RC filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty \frac{Q_f}{w} |H_r(jw)|^2 dw$$

$$H_r(s) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{1}{(1 + s\tau)}$$

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty \frac{Q_f}{w} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{1 + (w\tau)^2} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^\infty \frac{1}{w(1 + (w\tau)^2)} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left[ \frac{1}{2} \log \left( \frac{w^2}{1 + (w\tau)^2} \right) \right]_0^\infty$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left[ \frac{1}{2} \log \left( \frac{w^2}{1 + (w\tau)^2} \right) \right]_0^\infty$$

for a FET

$$\sigma_{V_{out}}^2(t_\infty) = \frac{Kf_2}{WLC_{ox}} \left( \frac{A c_{tot}}{c_{tot} + A c_f} \right)^2 \left[ \frac{1}{2} \log \left( \frac{w^2}{1 + (w\tau)^2} \right) \right]_0^\infty$$

So at infinite time after switch-on the noise is infinite.

B. Thermal noise through a charge amp and CR - RC filter in the time domain.

$$\sigma_{V_{out}}^2(t_{01}) = \int_{t=0}^{t=t_{01}} \pi Q_t h_r^2(t) dt$$

$$h_r(t_{m1} - t_p) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \left( \frac{1}{\tau} - \frac{t_{m1} - t_p}{(\tau)^2} \right) e^{-\frac{(t_{m1} - t_p)}{\tau}}$$

$$\sigma_{V_{out}}^2(t_{01}) = \pi Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^{t_{01}} \left( \frac{1}{\tau} - \frac{t}{\tau^2} \right)^2 e^{-\frac{2t}{\tau}} dt$$

$$\sigma_{V_{out}}^2(t_{01}) = \pi Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^{t_{01}} \left( \frac{1}{\tau^2} - \frac{2t}{\tau^3} + \frac{t^2}{\tau^4} \right) e^{-\frac{2t}{\tau}} dt$$

$$\sigma_{V_{out}}^2(t_{01}) = \pi Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left[ \left( \frac{-1}{2\tau} - 2 \left( -\frac{t\tau}{2} - \frac{\tau^2}{4} \right) \frac{1}{\tau^3} + \left( -\frac{t^2\tau}{2} - \frac{t\tau^2}{2} - \frac{\tau^3}{4} \right) \frac{1}{\tau^4} \right) e^{-\frac{2t}{\tau}} \right]_0^{t_{01}}$$

$$\sigma_{V_{out}}^2(t_{01}) = \pi Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left[ \left( \frac{-1}{2\tau} + \frac{t}{\tau^2} + \frac{1}{2\tau} - \frac{t^2}{2\tau^3} - \frac{t}{2\tau^2} - \frac{1}{4\tau} \right) e^{-\frac{2t}{\tau}} \right]_0^{t_{01}}$$

$$\sigma_{V_{out}}^2(t_{01}) = \pi Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left[ \left( \frac{t}{2\tau^2} - \frac{t^2}{2\tau^3} - \frac{1}{4\tau} \right) e^{-\frac{2t}{\tau}} \right]_0^{t_{01}}$$

$$\sigma_{V_{out}}^2(t_{01}) = \pi Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( \left( \frac{-1}{4\tau} + \frac{t_{01}}{2\tau^2} - \frac{t_{01}^2}{2\tau^3} \right) e^{-\frac{2t_{01}}{\tau}} + \frac{1}{4\tau} \right)$$

$$\sigma_{V_{out}}^2(t_{\infty}) = 2kTr \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \frac{1}{4\tau}$$

B. Thermal noise through a charge amp and CR - RC filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_{\infty}) = \int_0^{\infty} Q_i |H_r(jw)|^2 dw$$

$$H_r(s) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{s\tau}{(1+s\tau)^2}$$

$$\sigma_{V_{out}}^2(t_{\infty}) = \int_0^{\infty} Q_i |H_r(jw)|^2 dw$$

$$\sigma_{V_{out}}^2(t_{\infty}) = \int_0^{\infty} Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{(w\tau)^2}{(1+(w\tau)^2)^2} dw$$

$$\sigma_{V_{out}}^2(t_{\infty}) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \tau^2 \left( \left[ \frac{-w}{2\tau^2(1+(w\tau)^2)} \right]_0^{\infty} + \frac{1}{2\tau^2} \int_0^{\infty} \frac{dw}{1+(w\tau)^2} \right)$$

$$\sigma_{V_{out}}^2(t_{\infty}) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \tau^2 \left[ \frac{1}{2(\tau)^3} \tan^{-1}(w\tau) \right]_0^{\infty}$$

$$\sigma_{V_{out}}^2(t_{\infty}) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{\pi}{4\tau}$$

$$\sigma_{V_{out}}^2(t_{\infty}) = kTr \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \frac{1}{2\tau}$$

B. Shot noise through a charge amplifier and a CR-RC filter in the time domain.

$$\sigma_{V_{out}}^2(t_{01}) = \int_{t=0}^{t=t_{01}} \pi Q_s h_i^2(t) dt$$

$$h_i(t_{m1} - t_p) = \frac{A}{c_{tot} + A c_f} \frac{(t_{m1} - t_p)}{\tau} e^{-\frac{(t_{m1} - t_p)}{\tau}}$$

$$\sigma_{V_{out}}^2(t_{01}) = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{1}{\tau^2} \int_0^{t_{01}} t^2 e^{-\frac{2t}{\tau}} dt$$

$$\sigma_{V_{out}}^2(t_{01}) = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{1}{\tau^2} \left[ e^{-\frac{2t}{\tau}} \left( \frac{-t^2 \tau}{2} - \frac{\tau^2 t}{2} - \frac{2\tau^3}{8} \right) \right]_0^{t_{01}}$$

$$\sigma_{V_{out}}^2(t_{01}) = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{1}{\tau^2} \left( e^{-\frac{2t_{01}}{\tau}} \left( \frac{-t_{01}^2 \tau}{2} - \frac{\tau^2 t_{01}}{2} - \frac{2\tau^3}{8} \right) + \frac{2\tau^3}{8} \right)$$

$$\sigma_{V_{out}}^2(t_{\infty}) = iq \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{\tau}{4}$$

B. Shot noise through a charge amplifier and a CR-RC filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_s |H_i(jw)|^2 dw$$

$$H_i(s) = \frac{A}{c_{tot} + A c_f} \frac{\tau}{(1 + s\tau)^2}$$

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{\tau^2}{(1 + (w\tau)^2)^2} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \tau^2 \left( \left[ \frac{w}{2(1 + (w\tau)^2)} \right]_0^\infty + \frac{1}{2} \int_0^\infty \frac{dw}{1 + (w\tau)^2} \right)$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \tau^2 \left[ \frac{w}{2(1 + (w\tau)^2)} + \frac{1}{2\tau} \tan^{-1} w\tau \right]_0^\infty$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \tau^2 \frac{1}{2\tau} \frac{\pi}{2}$$

$$\sigma_{V_{out}}^2(t_\infty) = iq \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{\tau}{4}$$

B. Flicker noise through a charge amp and a CR-RC filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty \frac{Q_f}{w} |H_r(jw)|^2 dw$$

$$H_r(s) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{s\tau}{(1+s\tau)^2}$$

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty \frac{Q_f}{w} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{(w\tau)^2}{(1+(w\tau)^2)^2} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^\infty \frac{w\tau^2}{(1+(w\tau)^2)^2} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left[ \frac{-1}{2(1+(w\tau)^2)} \right]_0^\infty$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{2}$$

for a FET

$$\sigma_{V_{out}}^2(t_\infty) = \frac{Kf_2}{WLC_{ox}} \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \frac{1}{2}$$



B. Thermal noise through a charge amp and CR-RC<sup>2</sup>filter in the time domain.

$$\sigma_{V_{out}}^2(t_\infty) = \int_{t=0}^{t=\infty} Q_i h_r^2(t) dt$$

$$h_r(t) = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \left[ \frac{t}{\tau^2} - \frac{t^2}{2\tau^3} \right] e^{-\frac{t}{\tau}}$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^\infty \left[ \frac{t}{\tau^2} - \frac{t^2}{2\tau^3} \right]^2 e^{-\frac{2t}{\tau}} dt$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^\infty \left( \frac{t^2}{\tau^4} - \frac{t^3}{2\tau^5} + \frac{t^4}{4\tau^6} \right) e^{-\frac{2t}{\tau}} dt$$

ignoring the terms integrated to give  $t^n e^{-\frac{2t}{\tau}}$  as these will give zero in these limits

$$\sigma_{V_{out}}^2(t_\infty) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left[ \left[ \left( -\frac{\tau^3}{4} \right) \frac{1}{\tau^4} - \left( -\frac{3\tau^4}{4} \right) \frac{1}{2\tau^5} + \left( -\frac{3\tau^5}{4} \right) \frac{1}{4\tau^6} \right] e^{-\frac{2t}{\tau}} + \dots \text{ignored} \right]_0^\infty$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( \frac{1}{4\tau} - \frac{3}{8\tau} + \frac{3}{16\tau} \right)$$

$$\sigma_{V_{out}}^2(t_\infty) = kTr \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \frac{1}{8\tau}$$

B. Thermal noise through a charge amp and CR-RC<sup>2</sup> filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_t |H_r(jw)|^2 dw$$

$$H_r(s) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{s\tau}{(1+s\tau)^3}$$

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{(w\tau)^2}{(1+(w\tau)^2)^3} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( \left[ \frac{-w}{4\tau^2(1+(w\tau)^2)^2} \right]_0^\infty + \frac{1}{4\tau^2} \int_0^\infty \frac{dw}{(1+(w\tau)^2)^2} \right)$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( \left[ \frac{w}{2(1+(w\tau)^2)} \right]_0^\infty + \frac{1}{2} \int_0^\infty \frac{dw}{1+(w\tau)^2} \right)$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left[ \frac{1}{\tau} \tan^{-1}(w\tau) \right]_0^\infty$$

$$\sigma_{V_{out}}^2(t_\infty) = kTr \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \frac{1}{8\tau}$$

B. Shot noise through a charge amplifier and a CR-RC<sup>2</sup> filter in the time domain.

$$\sigma_{V_{out}}^2(t_{01}) = \int_{t=0}^{t=t_{01}} \pi Q_s h_i^2(t) dt$$

$$h_i(t_{m1} - t_p) = \frac{A}{c_{tot} + A c_f} \frac{(t_{m1} - t_p)^2}{2\tau^2} e^{-\frac{(t_{m1} - t_p)}{\tau}}$$

$$\sigma_{V_{out}}^2(t_{01}) = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{1}{4\tau^4} \int_0^{t_{01}} t^4 e^{-\frac{2t}{\tau}} dt$$

$$\sigma_{V_{out}}^2(t_{01}) = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{1}{4\tau^4} \left[ 24e^{-\frac{2t}{\tau}} \left( \frac{-\tau^5}{32} - \frac{\tau^4 t}{16} - \frac{\tau^3 t^2}{16} - \frac{\tau^2 t^3}{24} - \frac{\tau t^4}{48} \right) \right]_0^{t_{01}}$$

$$\sigma_{V_{out}}^2(t_{01}) = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{24}{4\tau^4} \left( e^{-\frac{2t_{01}}{\tau}} \left( \frac{-\tau^5}{32} - \frac{\tau^4 t_{01}}{16} - \frac{\tau^3 t_{01}^2}{16} - \frac{\tau^2 t_{01}^3}{24} - \frac{\tau t_{01}^4}{48} \right) + \frac{\tau^5}{32} \right)$$

$$\sigma_{V_{out}}^2(t_{\infty}) = iq \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{3\tau}{16}$$

B. Shot noise through a charge amplifier and a CR-RC<sup>2</sup> filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_s |H_i(jw)|^2 dw$$

$$H_i(s) = \frac{A}{c_{tot} + A c_f} \frac{\tau}{(1 + s\tau)^3}$$

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{\tau^2}{(1 + (w\tau)^2)^3} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \tau^2 \left( \left[ \frac{w}{4(1 + (w\tau)^2)^2} \right]_0^\infty + \frac{3}{4} \int_0^\infty \frac{dw}{(1 + (w\tau)^2)^2} \right)$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{3\tau^2}{4} \left( \left[ \frac{w}{2(1 + (w\tau)^2)} \right]_0^\infty + \frac{1}{2} \int_0^\infty \frac{dw}{(1 + (w\tau)^2)} \right)$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{3\tau^2}{4} \left[ \frac{w}{2(1 + (w\tau)^2)} + \frac{1}{2\tau} \tan^{-1} w\tau \right]_0^\infty$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{3\tau^2}{4} \frac{1}{2\tau} \frac{\pi}{2}$$

$$\sigma_{V_{out}}^2(t_\infty) = iq \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{3\tau}{16}$$

B. Flicker noise through a charge amp and CR-RC<sup>2</sup> filter.

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty \frac{Q_f}{w} |H_r(jw)|^2 dw$$

$$H_r(s) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{s\tau}{(1+s\tau)^3}$$

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty \frac{Q_f}{w} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{(w\tau)^2}{(1+(w\tau)^2)^3} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^\infty \frac{w\tau^2}{(1+(w\tau)^2)^3} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left[ \frac{-1}{4(1+(w\tau)^2)^2} \right]_0^\infty$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{4}$$

for a FET input

$$\sigma_{V_{out}}^2(t_\infty) = \frac{Kf_2}{WLC_{ox}} \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \frac{1}{4}$$

B. Thermal noise through a charge amp and CR-RC<sup>n</sup> filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_t |H_r(jw)|^2 dw$$

$$H_r(s) = \frac{A c_{tot} s \tau}{gm(c_{tot} + A c_f)(1 + s\tau)^{n+1}}$$

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{(w\tau)^2}{(1 + (w\tau)^2)^{n+1}} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^\infty \frac{(w\tau)^2}{(1 + (w\tau)^2)^{n+1}} d(w\tau)^2$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{2\tau} \int_0^\infty \frac{w\tau}{(1 + (w\tau)^2)^{n+1}} d(w\tau)^2$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 B\left(\frac{3}{2}, n - \frac{1}{2}\right)$$

$$\sigma_{V_{out}}^2(t_\infty) = \frac{kTr}{\pi\tau} \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 B\left(\frac{3}{2}, n - \frac{1}{2}\right)$$

B. Shot noise through a charge amplifier and a CR-RC<sup>n</sup> filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_s |H_i(jw)|^2 dw$$

$$H_i(s) = \frac{A}{c_{tot} + A c_f} \frac{\tau}{(1 + s\tau)^{n+1}}$$

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{\tau^2}{(1 + (w\tau)^2)^{n+1}} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_s \int_0^\infty \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{1}{(1 + (w\tau)^2)^{n+1}} \frac{1}{2w} d(w\tau)^2$$

$$\sigma_{V_{out}}^2(t_\infty) = \frac{Q_s}{2} \left( \frac{A}{c_{tot} + A c_f} \right)^2 \tau \int_0^\infty \frac{1}{(1 + (w\tau)^2)^{n+1}} \frac{1}{w\tau} d(w\tau)^2$$

$$\sigma_{V_{out}}^2(t_\infty) = \frac{iq\tau}{2\pi} \left( \frac{A}{c_{tot} + A c_f} \right)^2 B\left(\frac{1}{2}, n + \frac{1}{2}\right)$$

B. Flicker noise through a charge amp and a CR-RC<sup>n</sup> filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_{\infty}) = \int_0^{\infty} \frac{Q_f}{w} |H_r(jw)|^2 dw$$

$$H_r(s) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{s\tau}{(1+s\tau)^{n+1}}$$

$$\sigma_{V_{out}}^2(t_{\infty}) = \int_0^{\infty} \frac{Q_f}{w} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{(s\tau)^2}{(1+(w\tau)^2)^{n+1}} dw$$

$$\sigma_{V_{out}}^2(t_{\infty}) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^{\infty} \frac{(s\tau)^2}{(1+(w\tau)^2)^{n+1}} dw$$

$$\sigma_{V_{out}}^2(t_{\infty}) = \frac{Q_f}{2} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^{\infty} \frac{1}{(1+(w\tau)^2)^{n+1}} d(w\tau)^2$$

$$\sigma_{V_{out}}^2(t_{\infty}) = \frac{Q_f}{2} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{n}$$

For FET input

$$\sigma_{V_{out}}^2(t_{\infty}) = \frac{Kf_2}{2WLC_{ox}} \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \frac{1}{n}$$



B. Thermal noise through a charge amp and a CR<sup>2</sup>-RC bipolar filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_i |H_r(j\omega)|^2 d\omega$$

$$H_r(s) = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \frac{s^2 \tau^2}{(1 + s\tau)^3}$$

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{\omega^4 \tau^4}{(1 + (\omega\tau)^2)^3} d\omega$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \tau^4 \int_0^\infty \frac{\omega^4}{(1 + (\omega\tau)^2)^3} d\omega$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \tau^4 \left( \frac{1}{\tau^2} \int_0^\infty \frac{\omega^2}{(1 + (\omega\tau)^2)^2} d\omega - \frac{1}{\tau^2} \int_0^\infty \frac{\omega^2}{(1 + (\omega\tau)^2)^3} d\omega \right)$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \tau^2 \left( 0 + \left[ \frac{1}{2\tau^3} \tan^{-1} \omega\tau \right]_0^\infty - \left[ \frac{1}{8\tau^3} \tan^{-1} \omega\tau \right]_0^\infty \right)$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{3}{8\tau} \frac{\pi}{2}$$

$$\sigma_{V_{out}}^2(t_\infty) = \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \frac{3kTr}{8\tau}$$

B. Shot noise through a charge amplifier and a CR<sup>2</sup>-RC bipolar filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_s |H_i(jw)|^2 dw$$

$$H_i(s) = \frac{A}{c_{tot} + A c_f} \frac{s \tau^2}{(1 + s \tau)^3}$$

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{-w^2 \tau^4}{(1 + (w \tau)^2)^3} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \tau^4 \left( 0 + \left[ -\frac{1}{8 \tau^3} \tan^{-1} w \tau \right]_0^\infty \right)$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{\tau \pi}{16}$$

$$\sigma_{V_{out}}^2(t_\infty) = iq \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{\tau}{16}$$

B. Flicker noise through a charge amp and RC<sup>2</sup>-CR bipolar filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty \frac{Q_f}{w} |H_r(jw)|^2 dw$$

$$H_r(s) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{s^2 \tau^2}{(1 + s\tau)^3}$$

$$\sigma_{V_{out}}^2(t_\infty) = \int_0^\infty \frac{Q_f}{w} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{w^4 \tau^4}{(1 + (w\tau)^2)^3} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = \frac{Q_f}{w} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^\infty \frac{w^3 \tau^4}{(1 + (w\tau)^2)^3} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = \frac{Q_f}{w} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( \int_0^\infty \frac{w\tau^2}{(1 + (w\tau)^2)^2} dw - \int_0^\infty \frac{w\tau^2}{(1 + (w\tau)^2)^3} dw \right)$$

$$\sigma_{V_{out}}^2(t_\infty) = \frac{Q_f}{w} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( \left[ \frac{-1}{2(1 + (w\tau)^2)} \right]_0^\infty - \left[ \frac{-1}{4(1 + (w\tau)^2)^2} \right]_0^\infty \right)$$

$$\sigma_{V_{out}}^2(t_\infty) = \frac{Q_f}{w} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{4}$$

For FET input

$$\sigma_{V_{out}}^2(t_\infty) = \frac{Kf_2}{WLC_{ox}} \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \frac{1}{4}$$

B. Thermal noise through a charge amp and cusp filter in the time domain.

$$\begin{aligned}
\sigma_{V_{out}}^2(t_{01}) &= \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_t h_r^2(t_{m1} - t_p) dt_p \\
h_r(t_{m1} - t_p) &= \frac{1}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \left( 1 - e^{-\frac{t_s}{\tau}} \right) e^{-\frac{(t_{m1} - (t_p + t_s))}{\tau}} \quad t_0 \langle t_p \langle t_{m1} - t_s \\
h_r(t_{m1} - t_p) &= -\frac{1}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \left( e^{-\frac{(t_p + t_s - t_{m1})}{\tau}} - e^{-\frac{t_s}{\tau}} \right) \quad t_{m1} - t_s \langle t_p \langle t_{m1} \\
\sigma_{V_{out}}^2(t_{01}) &= \int_{t_p=t_0}^{t_p=t_{m1}-t_s} \frac{\pi Q_t}{\tau^2} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( 1 - e^{-\frac{t_s}{\tau}} \right)^2 e^{-\frac{2(t_{m1} - (t_p + t_s))}{\tau}} dt_p \\
&+ \int_{t_p=t_{m1}-t_s}^{t_p=t_{m1}} \frac{\pi Q_t}{\tau^2} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( e^{-\frac{2(t_p + t_s - t_{m1})}{\tau}} - 2e^{-\frac{(t_p + 2t_p - t_{m1})}{\tau}} + e^{-\frac{2t_s}{\tau}} \right) dt_p \\
\sigma_{V_{out}}^2(t_{01}) &= \frac{\pi Q_t}{\tau^2} \tau \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{2} \left[ \left( 1 - e^{-\frac{t_s}{\tau}} \right)^2 e^{-\frac{2(t_{m1} - (t_p + t_s))}{\tau}} \right]_{t_0}^{t_{m1}-t_s} \\
&+ \frac{\pi Q_t}{\tau^2} \tau \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left[ \left[ -\frac{1}{2} e^{-\frac{-2(t_p + t_s - t_{m1})}{\tau}} + 2e^{-\frac{(t_p + 2t_p - t_{m1})}{\tau}} + t_p e^{-\frac{2t_s}{\tau}} \right] \right]_{t_{m1}-t_s}^{t_{m1}} \\
\sigma_{V_{out}}^2(t_{01}) &= \frac{\pi Q_t}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{2} \left( 1 - e^{-\frac{t_s}{\tau}} \right)^2 \left( e^{-\frac{2(t_{m1} - (t_{m1} - t_s + t_s))}{\tau}} - e^{-\frac{2(t_{m1} - (t_0 + t_s))}{\tau}} \right) \\
&+ \frac{\pi Q_t}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \\
&\left( -\frac{1}{2} e^{-\frac{2(t_{m1} + t_s - t_{m1})}{\tau}} + 2e^{-\frac{(t_{m1} + 2t_s - t_{m1})}{\tau}} + t_{m1} e^{-\frac{2t_s}{\tau}} + \frac{1}{2} e^{-\frac{2(t_{m1} - t_s + t_s - t_{m1})}{\tau}} - 2e^{-\frac{(t_{m1} - t_s + 2t_s - t_{m1})}{\tau}} - (t_{m1} - t_s) e^{-\frac{2t_s}{\tau}} \right) \\
\sigma_{V_{out}}^2(t_{01}) &= \frac{\pi Q_t}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{2} \left( 1 - 2e^{-\frac{t_s}{\tau}} + e^{-\frac{2t_s}{\tau}} \right) \left( 1 - e^{-\frac{(t_{m1} - (t_0 + t_s))}{\tau}} \right) \\
&+ \frac{\pi Q_t}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( -\frac{1}{2} e^{-\frac{2t_s}{\tau}} + 2e^{-\frac{2t_s}{\tau}} + t_{m1} e^{-\frac{2t_s}{\tau}} + \frac{1}{2} - 2e^{-\frac{t_s}{\tau}} - (t_{m1} - t_s) e^{-\frac{2t_s}{\tau}} \right) \\
\sigma_{V_{out}}^2(t_{01}) &= \frac{\pi Q_t}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{2} \left( 1 - e^{-\frac{2(t_{m1} - (t_0 + t_s))}{\tau}} - 2e^{-\frac{t_s}{\tau}} + e^{-\frac{2(t_{m1} - (t_0 + t_s)) - 2t_s}{\tau}} + e^{-\frac{2t_s}{\tau}} - e^{-\frac{2(t_{m1} - t_0 - t_s) + 2t_s}{\tau}} \right)
\end{aligned}$$

B. Thermal noise through a charge amp and cusp filter in the time domain (cont).

$$\begin{aligned}
 & + \frac{\pi Q_t}{\tau} \left( \frac{A c_{tot}}{\text{gm}(c_{tot} + A c_f)} \right)^2 \left[ \left( \frac{1}{2} - 2e^{-\left(\frac{t_s}{\tau}\right)} + \left(\frac{3}{2} + t_s\right)e^{-\frac{2t_s}{\tau}} \right) \right] \\
 \sigma_{V_{out}}^2(t_{01}) &= \frac{\pi Q_t}{\tau} \left( \frac{A c_{tot}}{\text{gm}(c_{tot} + A c_f)} \right)^2 \left( \frac{1}{2} - \frac{1}{2}e^{-\frac{2(t_{m1}-t_0-t_s)}{\tau}} - e^{-\frac{t_s}{\tau}} + \frac{1}{2}e^{-\frac{2(t_{m1}-t_0)}{\tau}} + \frac{1}{2}e^{-\frac{2t_s}{\tau}} - \frac{1}{2}e^{-\frac{2(t_{m1}-t_0)}{\tau}} \right) \\
 & + \frac{\pi Q_t}{\tau} \left( \frac{A c_{tot}}{\text{gm}(c_{tot} + A c_f)} \right)^2 \left[ \left( \frac{1}{2} - 2e^{-\frac{t_s}{\tau}} + \left(\frac{3}{2} + t_s\right)e^{-\frac{2t_s}{\tau}} \right) \right] \\
 \sigma_{V_{out}}^2(t_{01}) &= \frac{\pi Q_t}{\tau} \left( \frac{A c_{tot}}{\text{gm}(c_{tot} + A c_f)} \right)^2 \left( 1 - \frac{1}{2}e^{-\frac{2(t_{m1}-t_0-t_s)}{\tau}} - \frac{5}{2}e^{-\frac{t_s}{\tau}} + (2+t_s)e^{-\frac{2t_s}{\tau}} \right)
 \end{aligned}$$

Let  $t_{m1} - t_0$  tend to infinity gives

$$\sigma_{V_{out}}^2(t_\infty) = \frac{\pi Q_t}{\tau} \left( \frac{A c_{tot}}{\text{gm}(c_{tot} + A c_f)} \right)^2 \left( 1 - \frac{5}{2}e^{-\frac{t_s}{\tau}} + (2+t_s)e^{-\frac{2t_s}{\tau}} \right)$$

Let  $t_s$  tend to infinity gives

$$\sigma_{V_{out}}^2(t_\infty) = \frac{2kTr}{\tau} \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2$$

B. Thermal noise through a charge amp and cusp filter in the frequency domain

$$\sigma_{V_{out}}^2(t_\infty) = \int_{w=0}^{w=\infty} Q(w) |H_r(jw)|^2 dw$$

$$\sigma_{V_{out}}^2(t_\infty) = Q_t \int_{w=0}^{w=\infty} |H_r(jw)|^2 dw$$

$$h_r(t) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{1}{\tau} e^{-\frac{|t|}{\tau}}$$

$$H_r(jw) = \mathfrak{F}(h_r(t))$$

$$H_r(jw) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{1}{\tau} \int_{-\infty}^{\infty} e^{-\frac{|t|}{\tau}} e^{-iwt} dt$$

$$H_r(jw) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{1}{\tau} \int_0^{\infty} e^{-\frac{t}{\tau}} e^{-iwt} dt + \frac{1}{\tau} \int_{-\infty}^0 e^{-\frac{t}{\tau}} e^{-iwt} dt$$

$$H_r(jw) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{1}{\tau} \int_0^{\infty} e^{-t(\frac{1}{\tau} + iw)} dt + \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{1}{\tau} \int_{-\infty}^0 e^{-t(\frac{-1}{\tau} + iw)} dt$$

$$H_r(jw) = -\frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{1}{\tau} \left[ \frac{-e^{-t(\frac{1}{\tau} + iw)}}{\left(\frac{1}{\tau} + iw\right)} \right]_0^{\infty} - \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{1}{\tau} \left[ \frac{e^{-t(\frac{-1}{\tau} + iw)}}{\left(\frac{-1}{\tau} + iw\right)} \right]_{-\infty}^0$$

$$H_r(jw) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \left( \frac{1}{\tau \left(\frac{1}{\tau} + iw\right)} - \frac{1}{\tau \left(\frac{1}{\tau} - iw\right)} \right)$$

$$H_r(jw) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{-2iw}{\tau \left(\frac{1}{\tau^2} + w^2\right)} = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{-2iw\tau}{1 + (w\tau)^2}$$

$$\sigma_{V_{out}}^2(t_\infty) = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{4Q_t}{\tau} \int_{w=0}^{w=\infty} \frac{w^2 \tau^2}{(1 + (w\tau)^2)^2} d(w\tau)$$

$$\sigma_{V_{out}}^2(t_\infty) = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{4Q_t}{\tau} \left[ \frac{-w\tau}{2(1 + (w\tau)^2)} - \frac{1}{2} \tan^{-1} w\tau \right]_0^{\infty}$$

$$\sigma_{V_{out}}^2(t_\infty) = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{4Q_t}{\tau} \left[ \frac{\pi}{4} \right] = \left( \frac{A c_{tot}}{c_{tot} + A c_f} \right)^2 \frac{2kTr}{\tau}$$

B. Shot noise through a charge amp and cusp filter in the time domain.

$$\begin{aligned} \sigma_{V_{out}}^2(t_{01}) &= \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_s h_i^2(t_{m1} - t_p) dt_p \\ h_i(t_{m1} - t_p) &= \frac{A}{c_{tot} + A c_f} \left(1 - e^{-\frac{t_s}{\tau}}\right) e^{-\frac{(t_{m1} - (t_p + t_s))}{\tau}} \quad t_0 \langle t_p \rangle t_{m1} - t_s \\ h_i(t_{m1} - t_p) &= \frac{A}{c_{tot} + A c_f} \left( e^{-\frac{(t_p + t_s - t_{m1})}{\tau}} - e^{-\frac{t_s}{\tau}} \right) \quad t_{m1} - t_s \langle t_p \rangle t_{m1} \\ \sigma_{V_{out}}^2(t_{01}) &= \int_{t_p=t_0}^{t_p=t_{m1}-t_s} \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left(1 - e^{-\frac{t_s}{\tau}}\right)^2 e^{-\frac{2(t_{m1} - (t_p + t_s))}{\tau}} dt_p \\ &+ \int_{t_p=t_{m1}-t_s}^{t_p=t_{m1}} \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( e^{-\frac{2(t_p + t_s - t_{m1})}{\tau}} - 2e^{-\frac{(t_p + 2t_p - t_{m1})}{\tau}} + e^{-\frac{2t_s}{\tau}} \right) dt_p \\ \sigma_{V_{out}}^2(t_{01}) &= \pi Q_s \tau \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{1}{2} \left[ \left(1 - e^{-\frac{t_s}{\tau}}\right)^2 e^{-\frac{2(t_{m1} - (t_p + t_s))}{\tau}} \right]_{t_0}^{t_{m1}-t_s} \\ &+ \pi Q_s \tau \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left[ \left[ -\frac{1}{2} e^{-\frac{2(t_p + t_s - t_{m1})}{\tau}} + 2e^{-\frac{(t_p + 2t_p - t_{m1})}{\tau}} + t_p e^{-\frac{2t_s}{\tau}} \right] \right]_{t_{m1}-t_s}^{t_{m1}} \\ \sigma_{V_{out}}^2(t_{01}) &= \pi Q_s \tau \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{1}{2} \left(1 - e^{-\frac{t_s}{\tau}}\right)^2 \left( e^{-\frac{2(t_{m1} - (t_{m1} - t_s + t_s))}{\tau}} - e^{-\frac{2(t_{m1} - (t_0 + t_s))}{\tau}} \right) \\ &+ \pi Q_s \tau \left( \frac{A}{c_{tot} + A c_f} \right)^2 \\ &\left( -\frac{1}{2} e^{-\frac{2(t_{m1} + t_s - t_{m1})}{\tau}} + 2e^{-\frac{(t_{m1} + 2t_s - t_{m1})}{\tau}} + t_{m1} e^{-\frac{2t_s}{\tau}} + \frac{1}{2} e^{-\frac{2(t_{m1} - t_s + t_s - t_{m1})}{\tau}} - 2e^{-\frac{(t_{m1} - t_s + 2t_s - t_{m1})}{\tau}} - (t_{m1} - t_s) e^{-\frac{2t_s}{\tau}} \right) \\ \sigma_{V_{out}}^2(t_{01}) &= \pi Q_s \tau \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{1}{2} \left(1 - 2e^{-\frac{t_s}{\tau}} + e^{-\frac{2t_s}{\tau}}\right) \left(1 - e^{-\frac{2(t_{m1} - (t_0 + t_s))}{\tau}}\right) \\ &+ \pi Q_s \tau \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( -\frac{1}{2} e^{-\frac{2t_s}{\tau}} + 2e^{-\frac{2t_s}{\tau}} + t_{m1} e^{-\frac{2t_s}{\tau}} + \frac{1}{2} - 2e^{-\frac{t_s}{\tau}} - (t_{m1} - t_s) e^{-\frac{2t_s}{\tau}} \right) \end{aligned}$$

B. Shot noise through a charge amp and cusp filter in the time domain (cont)

$$\sigma_{Vout}^2(t_{01}) = \pi Q_s \tau \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{1}{2} \left( 1 - e^{-\frac{2(t_{m1} - (t_0 + t_s))}{\tau}} - 2e^{-\frac{t_s}{\tau}} + e^{-\frac{2(t_{m1} - (t_0 + t_s)) + t_s}{\tau}} + e^{-\frac{2t_s}{\tau}} - e^{-\frac{2(t_{m1} - t_0 - t_s) + 2t_s}{\tau}} \right)$$

$$+ \pi Q_s \tau \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left[ \left( \frac{1}{2} - 2e^{-\frac{t_s}{\tau}} + \left( \frac{3}{2} + t_s \right) e^{-\frac{2t_s}{\tau}} \right) \right]$$

$$\sigma_{Vout}^2(t_{01}) = \pi Q_s \tau \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( \frac{1}{2} - \frac{1}{2} e^{-\frac{2(t_{m1} - t_0 - t_s)}{\tau}} - e^{-\frac{t_s}{\tau}} + \frac{1}{2} e^{-\frac{2(t_{m1} - t_0)}{\tau}} + \frac{1}{2} e^{-\frac{2t_s}{\tau}} - \frac{1}{2} e^{-\frac{2(t_{m1} - t_0 + t_s)}{\tau}} \right)$$

$$+ \pi Q_s \tau \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left[ \left( \frac{1}{2} - 2e^{-\frac{t_s}{\tau}} + \left( \frac{3}{2} + t_s \right) e^{-\frac{2t_s}{\tau}} \right) \right]$$

$$\sigma_{Vout}^2(t_{01}) = \pi Q_s \tau \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( 1 - \frac{1}{2} e^{-\frac{2(t_{m1} - t_0 - t_s)}{\tau}} - \frac{5}{2} e^{-\frac{t_s}{\tau}} + \frac{1}{2} e^{-\frac{(t_{m1} - t_0)}{\tau}} + (2 + t_s) e^{-\frac{2t_s}{\tau}} - \frac{1}{2} e^{-\frac{(t_{m1} - t_0)}{\tau}} \right)$$

Letting  $t_{m1} - t_0$  tend to infinity gives

$$\sigma_{Vout}^2(t_\infty) = i q \tau \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( 1 - \frac{5}{2} e^{-\frac{t_s}{\tau}} + (2 + t_s) e^{-\frac{2t_s}{\tau}} \right)$$

Letting  $t_s$  tend to infinity gives

$$\sigma_{Vout}^2(t_\infty) = i q \tau \left( \frac{A}{c_{tot} + A c_f} \right)^2$$



B. Shot noise through a charge amp and cusp filter in the frequency domain.

$$\sigma_{Vout}^2(t_\infty) = \int_{w=0}^{w=\infty} Q_s |H_i(jw)|^2 dw$$

$$\sigma_{Vout}^2(t_\infty) = Q_s \int_{w=0}^{w=\infty} |H_i(jw)|^2 dw$$

$$h_i(t) = \frac{A}{c_{tot} + A c_f} e^{-\frac{|t|}{\tau}}$$

$$H_i(jw) = \mathfrak{F}(h_i(t))$$

$$H_i(jw) = \frac{A}{c_{tot} + A c_f} \int_{-\infty}^{\infty} e^{-\frac{|t|}{\tau}} e^{-j\omega t} dt$$

$$H_i(jw) = \frac{A}{c_{tot} + A c_f} \int_0^{\infty} e^{-\frac{t}{\tau}} e^{-j\omega t} dt + \frac{A}{c_{tot} + A c_f} \int_{-\infty}^0 e^{-\frac{|t|}{\tau}} e^{-j\omega t} dt$$

$$H_i(jw) = \frac{A}{c_{tot} + A c_f} \int_0^{\infty} e^{-t(\frac{1}{\tau} + j\omega)} dt + \frac{A}{c_{tot} + A c_f} \int_{-\infty}^0 e^{-t(\frac{-1}{\tau} + j\omega)} dt$$

$$H_i(jw) = \frac{A}{c_{tot} + A c_f} \left[ \frac{e^{-t(\frac{1}{\tau} + j\omega)}}{-\left(\frac{1}{\tau} + j\omega\right)} \right]_0^{\infty} + \frac{A}{c_{tot} + A c_f} \left[ \frac{e^{-t(\frac{-1}{\tau} + j\omega)}}{-\left(\frac{-1}{\tau} + j\omega\right)} \right]_{-\infty}^0$$

$$H_i(jw) = \frac{A}{c_{tot} + A c_f} \left( \frac{1}{\left(\frac{1}{\tau} + j\omega\right)} + \frac{1}{\left(\frac{1}{\tau} - j\omega\right)} \right)$$

$$H_i(jw) = \frac{A}{c_{tot} + A c_f} \frac{2}{\tau \left( \frac{1}{\tau^2} + \omega^2 \right)}$$

$$\sigma_{Vout}^2(t_\infty) = \left( \frac{A}{c_{tot} + A c_f} \right)^2 Q_s \frac{4}{\tau^2} \int_{w=0}^{w=\infty} \frac{1}{\left( \frac{1}{\tau^2} + \omega^2 \right)^2} d\omega$$

$$\sigma_{Vout}^2(t_\infty) = \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{Q_s 4\tau^2}{2\tau^2} \left[ \frac{w}{\left( \frac{1}{\tau^2} + \omega^2 \right)} + \tau \tan^{-1} \omega \tau \right]_0^{\infty}$$

$$\sigma_{Vout}^2(t_\infty) = \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{Q_s 4\tau^2}{2\tau^2} \frac{\pi}{2}$$

$$\sigma_{Vout}^2(t_\infty) = \left( \frac{A}{c_{tot} + A c_f} \right)^2 i q \tau$$

B. Flicker noise through a charge amp and cusp filter in the frequency domain.

$$\sigma_{V_{out}}^2(t_\infty) = \int_{w=0}^{w=\infty} \frac{Q_f}{w} |H_r(jw)|^2 dw$$

$$h_r(t) = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \frac{-1}{\tau} e^{-\frac{|t|}{\tau}}$$

$$H_r(jw) = \mathfrak{F}(h(t))$$

$$H_r(jw) = \frac{-1}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \int_{-\infty}^{\infty} e^{-\frac{|t|}{\tau}} e^{-iwt} dt$$

$$H_r(jw) = \frac{-1}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \int_0^{\infty} e^{-\frac{t}{\tau}} e^{-iwt} dt - \frac{1}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \int_{-\infty}^0 e^{\frac{t}{\tau}} e^{-iwt} dt$$

$$H_r(jw) = \frac{-1}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \int_0^{\infty} e^{-t \left( \frac{1}{\tau} + iw \right)} dt - \frac{1}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \int_{-\infty}^0 e^{-t \left( \frac{1}{\tau} + iw \right)} dt$$

$$H_r(jw) = \frac{-1}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \left[ \frac{e^{-t \left( \frac{1}{\tau} + iw \right)}}{\left( \frac{1}{\tau} + iw \right)} \right]_0^{\infty} - \frac{1}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \left[ \frac{e^{-t \left( \frac{1}{\tau} + iw \right)}}{\left( \frac{1}{\tau} + iw \right)} \right]_{-\infty}^0$$

$$H_r(jw) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \left( \frac{1}{\tau \left( \frac{1}{\tau} + iw \right)} - \frac{1}{\tau \left( \frac{1}{\tau} - iw \right)} \right)$$

$$H_r(jw) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{2iw}{\tau \left( \frac{1}{\tau^2} + w^2 \right)}$$

$$\sigma_{V_{out}}^2(t_\infty) = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{-4Q_f}{\tau^2} \int_{w=0}^{w=\infty} \frac{w}{\left( \frac{1}{\tau^2} + w^2 \right)^2} dw$$

$$\sigma_{V_{out}}^2(t_\infty) = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{-4Q_f}{\tau^2} \left[ \frac{1}{2 \left( \frac{1}{\tau^2} + w^2 \right)} \right]_0^{\infty}$$

$$\sigma_{V_{out}}^2(t_\infty) = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{4Q_f}{\tau^2} \frac{\tau^2}{2} = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 2Q_f$$

!For FET input

$$\sigma_{V_{out}}^2(t_\infty) = \left( \frac{A c_{tot}}{c_{tot} + A c_f} \right)^2 \frac{2Kf_2}{WLC_{ax}}$$

B. Thermal noise through a charge amp and a trapezoidal filter in the time domain.

$$\begin{aligned}
 \sigma_{Vout}^2(t_{01}) &= \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_i h_r^2(t_m - t_p) dt_p \\
 h_r(t_{m1} - t_p) &= 0 & t_0 < t_p < t_{m1} - (t_{rise} + t_{flat} + t_{fall}) \\
 h_r(t_{m1} - t_p) &= \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{1}{t_{fall}} & t_{m1} - (t_{rise} + t_{flat} + t_{fall}) < t_p < t_{m1} - (t_{rise} + t_{flat}) \\
 h_r(t_{m1} - t_p) &= 0 & t_{m1} - (t_{rise} + t_{flat}) < t_p < t_{m1} - t_{rise} \\
 h_r(t_{m1} - t_p) &= \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{-1}{t_{rise}} & t_{m1} - t_{rise} < t_p < t_{m1}
 \end{aligned}$$

$$\begin{aligned}
 \frac{\sigma_{Vout}^2(t_{01})}{\left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2} &= \int_{t_p=t_{m1}-(t_{rise}+t_{flat}+t_{fall})}^{t_p=t_{m1}-(t_{rise}+t_{flat})} \left( \frac{1}{t_{fall}} \right)^2 dt_p + \int_{t_p=t_{m1}-t_{rise}}^{t_p=t_{m1}} \left( \frac{-1}{t_{rise}} \right)^2 dt_p \\
 &= \frac{1}{t_{fall}^2} \left[ t_p \right]_{t_{m1}-(t_{rise}+t_{flat}+t_{fall})}^{t_{m1}-(t_{rise}+t_{flat})} + \frac{1}{t_{rise}^2} \left[ t_p \right]_{t_{m1}-t_{rise}}^{t_{m1}} \\
 &= \frac{1}{t_{fall}^2} (t_{m1} - t_{rise} - t_{flat}) - \frac{1}{t_{fall}^2} (t_{m1} - t_{rise} - t_{flat} - t_{fall}) + \frac{1}{t_{rise}^2} t_{m1} - \frac{1}{t_{rise}^2} (t_{m1} - t_{rise}) \\
 \sigma_{Vout}^2(t_{01}) &= 2kTr \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \left( \frac{1}{t_{fall}} + \frac{1}{t_{rise}} \right)
 \end{aligned}$$

B. Thermal noise through charge amp and a trapezoidal filter in the frequency domain.

Transfer function of trapezoidal step response with  $t_{rise} = t_{fall}$

$$H_v(s) = \frac{1}{t_{rise}} \left( \frac{1}{s^2} - \frac{e^{st_{rise}}}{s^2} - \frac{e^{s(t_{rise}+t_{flat})}}{s^2} + \frac{e^{s(2t_{rise}+t_{flat})}}{s^2} \right)$$

$$|H_v(j\omega)|^2 =$$

$$\frac{1}{\omega^4 t_{rise}^2} \left( 1 - e^{j\omega t_{rise}} - e^{j\omega(t_{rise}+t_{flat})} + e^{j\omega(2t_{rise}+t_{flat})} \right) \left( 1 - e^{-j\omega t_{rise}} - e^{-j\omega(t_{rise}+t_{flat})} + e^{-j\omega(2t_{rise}+t_{flat})} \right)$$

$$|H_v(j\omega)|^2 \omega^4 t_{rise}^2 =$$

$$1 - e^{-j\omega t_{rise}} - e^{-j\omega(t_{rise}+t_{flat})} + e^{-j\omega(2t_{rise}+t_{flat})} - e^{j\omega t_{rise}} + 1 + e^{-j\omega t_{flat}} - e^{-j\omega(t_{rise}+t_{flat})} \\ - e^{j\omega(t_{rise}+t_{flat})} + e^{j\omega t_{flat}} + 1 - e^{-j\omega t_{rise}} + e^{j\omega(2t_{rise}+t_{flat})} - e^{j\omega(t_{rise}+t_{flat})} - e^{j\omega t_{rise}} + 1$$

$$|H_v(j\omega)|^2 \omega^4 t_{rise}^2 = 4 - 2e^{-j\omega t_{rise}} - 2e^{j\omega t_{rise}} - 2e^{-j\omega(t_{rise}+t_{flat})} - 2e^{j\omega(t_{rise}+t_{flat})} + e^{-j\omega t_{flat}} + e^{j\omega t_{flat}} \\ + e^{-j\omega(2t_{rise}+t_{flat})} + e^{j\omega(2t_{rise}+t_{flat})}$$

$$|H_v(j\omega)|^2 \omega^4 t_{rise}^2 = 4 - 4 \cos \omega t_{rise} - 4 \cos \omega(t_{rise} + t_{flat}) + 2 \cos \omega t_{flat} + 2 \cos \omega(2t_{rise} + t_{flat})$$

$$|H_v(j\omega)|^2 \omega^4 t_{rise}^2 = 4 - 4 \cos \omega t_{rise} - 4 \cos \omega(t_{rise} + t_{flat}) + 4 \cos \omega(t_{rise} + t_{flat}) \cos \omega(t_{rise})$$

$$|H_v(j\omega)|^2 \omega^4 t_{rise}^2 = 4 - 4 \cos \omega t_{rise} - 4(1 - \cos \omega t_{rise}) \cos \omega(t_{rise} + t_{flat})$$

$$|H_v(j\omega)|^2 \omega^4 t_{rise}^2 = 4(1 - \cos \omega t_{rise}) - 4(1 - \cos \omega t_{rise}) \cos \omega(t_{rise} + t_{flat})$$

$$|H_v(j\omega)|^2 \omega^4 t_{rise}^2 = 4(1 - \cos \omega t_{rise})(1 - \cos \omega(t_{rise} + t_{flat}))$$

$$|H_v(j\omega)|^2 \omega^4 t_{rise}^2 = 4(2 \sin^2 \omega t_{rise})(2 \sin^2 \omega(t_{rise} + t_{flat}))$$

$$|H_v(j\omega)|^2 = \frac{16 \sin^2 \frac{\omega t_{rise}}{2} \sin^2 \left( \frac{t_{rise} + t_{flat}}{2} \right)}{\omega^4 t_{rise}^2}$$

Transfer function of trapezoidal impulse response

$$|H_v(j\omega)|^2 = \frac{16 \sin^2 \frac{\omega t_{rise}}{2} \sin^2 \left( \frac{t_{rise} + t_{flat}}{2} \right)}{\omega^2 t_{rise}^2}$$

B. Thermal noise in charge amp and trapezoidal filter in the frequency domain (cont).

$$\sigma_{Vout}^2(t_\infty) = \int_{w=0}^{w=\infty} Q_t |H_r(jw)|^2 dw$$

$$|H_r(jw)|^2 = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{16}{w^2 t_{rise}^2} \sin^2(wt_{rise}/2) \sin^2(w(t_{rise} + t_{flat})/2)$$

$$\sigma_{Vout}^2(t_\infty) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_{w=0}^{w=\infty} \frac{16}{w^2 t_{rise}^2} \sin^2(wt_{rise}/2) \sin^2(w(t_{rise} + t_{flat})/2) dw$$

if  $t_{flat} = 0$

$$\sigma_{Vout}^2(t_\infty) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_{w=0}^{w=\infty} \frac{16}{w^2 t_{rise}^2} \sin^4(wt_{rise}/2) dw$$

$$\sigma_{Vout}^2(t_\infty) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{16}{2t_{rise}} \int_{w=0}^{w=\infty} \frac{4}{w^2 t_{rise}^2} \sin^4(wt_{rise}/2) d(wt_{rise}/2)$$

$$\sigma_{Vout}^2(t_\infty) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{16}{2t_{rise}} \frac{\pi}{4} = \frac{4kTr}{t_{rise}} \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2$$

if  $t_{flat} = t_{rise}$

$$\sigma_{Vout}^2(t_\infty) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_{w=0}^{w=\infty} \frac{16}{w^2 t_{rise}^2} \sin^2(wt_{rise}/2) \sin^2(wt_{rise}) dw$$

$$\sigma_{Vout}^2(t_\infty) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{16}{2t_{rise}} \int_{w=0}^{w=\infty} \frac{4}{w^2 t_{rise}^2} \sin^2(wt_{rise}/2) \sin^2(wt_{rise}) d(wt_{rise}/2)$$

$$\sigma_{Vout}^2(t_\infty) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{8}{t_{rise}} \frac{\pi}{4} = \frac{4kTr}{t_{rise}} \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2$$

B. Shot noise through charge amp and a trapezoidal filter in the time domain.

$$\sigma_{V_{out}}^2(t_{01}) = \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_s h_i^2(t_m - t_p) dt_p$$

$$h_i(t_m - t_p) = 0$$

$$t_0 \langle t_p \langle t_{m1} - (t_{rise} + t_{flat} + t_{fall}) \rangle \rangle$$

$$h_i(t_m - t_p) = \frac{A}{c_{tot} + A c_f} \frac{(t_p + t_{rise} + t_{flat} + t_{fall}) - t_{m1}}{t_{fall}}$$

$$t_{m1} - (t_{rise} + t_{flat} + t_{fall}) \langle t_p \langle t_{m1} - (t_{rise} + t_{flat}) \rangle \rangle$$

$$h_i(t_m - t_p) = 1$$

$$t_{m1} - (t_{rise} + t_{flat}) \langle t_p \langle t_{m1} - t_{rise} \rangle \rangle$$

$$h_i(t_m - t_p) = \frac{A}{c_{tot} + A c_f} \frac{t_{m1} - t_p}{t_{rise}}$$

$$t_{m1} - t_{rise} \langle t_p \langle t_{m1} \rangle \rangle$$

$$\frac{\sigma_{V_{out}}^2(t_{01})}{\pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2} =$$

$$= \int_{t_p=t_{m1}-(t_{rise}+t_{flat})}^{t_p=t_{m1}-(t_{rise}+t_{flat})} \left( \frac{(t_p + t_{rise} + t_{flat} + t_{fall}) - t_{m1}}{t_{fall}} \right)^2 dt_p + \int_{t_p=t_{m1}-(t_{rise}+t_{flat})}^{t_p=t_{m1}-t_{rise}} dt_p + \int_{t_p=t_{m1}-t_{rise}}^{t_p=t_{m1}} \left( \frac{t_{m1} - t_p}{t_{rise}} \right)^2 dt_p$$

substituting in first integral  $t = t_{m1} - t_{rise} - t_{flat}$

$$= \frac{1}{t_{fall}^2} \int_{t_p=t-t_{fall}}^{t_p=t} (t_p - t + t_{fall})^2 dt_p + \int_{t_p=t_{m1}-(t_{rise}+t_{flat})}^{t_p=t_{m1}-t_{rise}} dt_p + \frac{1}{t_{rise}^2} \int_{t_p=t_{m1}-t_{rise}}^{t_p=t_{m1}} (t_{m1} - t_p)^2 dt_p$$

substituting in first integral  $t = t_p - t$  and in third integral  $t = t_p - t_{m1}$

$$\begin{aligned} &= \frac{1}{t_{fall}^2} \int_{t=-t_{fall}}^{t=0} (t + t_{fall})^2 dt + \int_{t_p=t_{m1}-(t_{rise}+t_{flat})}^{t_p=t_{m1}-t_{rise}} dt_p + \frac{1}{t_{rise}^2} \int_{t=-t_{rise}}^{t=0} t^2 dt \\ &= \frac{1}{t_{fall}^2} \int_{t=-t_{fall}}^{t=0} (t^2 + 2t_{fall}t + t_{fall}^2) dt + \int_{t_p=t_{m1}-(t_{rise}+t_{flat})}^{t_p=t_{m1}-t_{rise}} dt_p + \frac{1}{t_{rise}^2} \int_{t=-t_{rise}}^{t=0} t^2 dt_p \\ &= \frac{1}{t_{fall}^2} \left[ \frac{t^3}{3} + t_{fall}t^2 + tt_{fall}^2 \right]_{-t_{fall}}^0 + [t_p]_{t_{m1}-(t_{rise}+t_{flat})}^{t_{m1}-t_{rise}} + \frac{1}{t_{rise}^2} \left[ \frac{t^3}{3} \right]_{-t_{rise}}^0 \\ &= \frac{1}{t_{fall}^2} \left( \frac{t_{fall}^3}{3} + t_{fall}t_{fall}^2 - t_{fall}t_{fall}^2 \right) + t_{m1} - t_{rise} - t_{m1} + (t_{rise} + t_{flat}) - \frac{(-t_{rise})^3}{3t_{rise}^2} \\ &= \frac{t_{fall}}{3} + t_{flat} + \frac{t_{rise}}{3} \end{aligned}$$

$$\sigma_{V_{out}}^2(t_{01}) = iQ \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( \frac{t_{fall}}{3} + t_{flat} + \frac{t_{rise}}{3} \right)$$

B. Shot noise in charge amp and trapezoidal filter in the frequency domain.

$$\sigma_{Vout}^2(t_\infty) = \int_{w=0}^{w=\infty} Q_s |H_i(jw)|^2 dw$$

$$|H_i(jw)|^2 = \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{16}{w^4 t_{rise}^2} \sin^2(wt_{rise}/2) \sin^2(w(t_{rise} + t_{flat})/2)$$

$$\sigma_{Vout}^2(t_\infty) = Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \int_{w=0}^{w=\infty} \frac{16}{w^4 t_{rise}^2} \sin^2(wt_{rise}/2) \sin^2(w(t_{rise} + t_{flat})/2) dw$$

if  $t_{flat} = 0$

$$\sigma_{Vout}^2(t_\infty) = Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \int_{w=0}^{w=\infty} \frac{16}{w^4 t_{rise}^2} \sin^4(wt_{rise}/2) dw$$

$$\sigma_{Vout}^2(t_\infty) = Q_s 16 \left( \frac{A}{c_{tot} + A c_f} \right)^2 \int_{w=0}^{w=\infty} \frac{2}{w^4 t_{rise}^3} \sin^4(wt_{rise}/2) d(wt_{rise}/2)$$

$$\sigma_{Vout}^2(t_\infty) = Q_s 16 \frac{t_{rise}}{8} \left( \frac{A}{c_{tot} + A c_f} \right)^2 \int_{w=0}^{w=\infty} \frac{16}{w^4 t_{rise}^4} \sin^4(wt_{rise}/2) d(wt_{rise}/2)$$

$$\sigma_{Vout}^2(t_\infty) = Q_s 2 \frac{\pi}{3} \left( \frac{A}{c_{tot} + A c_f} \right)^2 t_{rise} = \frac{2qi}{3} \left( \frac{A}{c_{tot} + A c_f} \right)^2 t_{rise}$$

if  $t_{flat} = t_{rise}$

$$\sigma_{Vout}^2(t_\infty) = Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \int_{w=0}^{w=\infty} \frac{16}{w^4 t_{rise}^2} \sin^2(wt_{rise}/2) \sin^2(wt_{rise}) dw$$

$$\sigma_{Vout}^2(t_\infty) = Q_s 2t_{rise} \left( \frac{A}{c_{tot} + A c_f} \right)^2 \int_{w=0}^{w=\infty} \frac{16}{w^4 t_{rise}^4} \sin^2(wt_{rise}/2) \sin^2(wt_{rise}) d(wt_{rise}/2)$$

$$\sigma_{Vout}^2(t_\infty) = Q_s 2t_{rise} \left( \frac{A}{c_{tot} + A c_f} \right)^2 \int_{x=0}^{x=\infty} \frac{1}{x^4} \sin^2 x \sin^2 2x dx$$

$$\sigma_{Vout}^2(t_\infty) = Q_s 2 \frac{5\pi}{6} t_{rise} \left( \frac{A}{c_{tot} + A c_f} \right)^2 = \frac{5qi}{3} \left( \frac{A}{c_{tot} + A c_f} \right)^2 t_{rise}$$

B. Flicker noise in charge amp and trapezoidal filter in the frequency domain.

$$\sigma_{Vout}^2(t_\infty) = \int_{w=0}^{w=\infty} \frac{Q_f}{w} |H_r(jw)|^2 dw$$

$$|H_r(jw)|^2 = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{16}{w^3 t_{rise}^2} \sin^2(wt_{rise}/2) \sin^2(w(t_{rise} + t_{flat})/2)$$

$$\sigma_{Vout}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_{w=0}^{w=\infty} \frac{16}{w^3 t_{rise}^2} \sin^2(wt_{rise}/2) \sin^2(w(t_{rise} + t_{flat})/2) dw$$

if  $t_{flat} = 0$

$$\sigma_{Vout}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_{w=0}^{w=\infty} \frac{16}{w^3 t_{rise}^2} \sin^4(wt_{rise}/2) dw$$

$$\sigma_{Vout}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{16}{4} \int_{w=0}^{w=\infty} \frac{8}{w^3 t_{rise}^3} \sin^4(wt_{rise}/2) d(wt_{rise}/2)$$

$$\sigma_{Vout}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{16}{4} \ln 2 = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 4 \ln 2$$

$$\sigma_{Vout}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 2.77$$

$$\sigma_{Vout}^2(t_\infty) = \frac{Kf_2}{WLC_{ox}} \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 2.77$$

if  $t_{flat} = t_{rise}$

$$\sigma_{Vout}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_{w=0}^{w=\infty} \frac{16}{w^3 t_{rise}^2} \sin^2(wt_{rise}/2) \sin^2(wt_{rise}) dw$$

$$\sigma_{Vout}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{16}{4} \int_{w=0}^{w=\infty} \frac{8}{w^3 t_{rise}^3} \sin^2(wt_{rise}/2) \sin^2(wt_{rise}) d(wt_{rise}/2)$$

$$\sigma_{Vout}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{16}{4} \ln \frac{9}{4} 3^{\frac{1}{4}} = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 4.34$$

$$\sigma_{Vout}^2(t_\infty) = \frac{Kf_2}{WLC_{ox}} \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 4.34$$



## B. Thermal noise through charge amp, RC filter and a

Correlated Double Sampler in the time domain.

$$\begin{aligned} \sigma_{sum}^2 &= \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_i \left( -h_r(t_{m1} - t_p) + h_r(t_{m2} - t_p) \right)^2 dt_p + \int_{t_p=t_{m1}}^{t_p=t_{m2}} \pi Q_i \left( h_r(t_{m2} - t_p) \right)^2 dt_p \\ h_r(t_m - t_p) &= \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right) \frac{1}{\tau} e^{-\frac{t_m - t_p}{\tau}} \\ \sigma_{sum}^2 &= \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{\tau^2} \left( e^{-\frac{t_{m2} - t_p}{\tau}} - e^{-\frac{t_{m1} - t_p}{\tau}} \right)^2 dt_p \\ &\quad + \int_{t_p=t_{m1}}^{t_p=t_{m2}} \pi Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{1}{\tau^2} \left( e^{-\frac{t_{m2} - t_p}{\tau}} \right)^2 dt_p \\ \sigma_{sum}^2 &= \frac{\pi Q_i}{\tau^2} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( \int_{t_p=t_0}^{t_p=t_{m1}} \left( e^{-2\frac{t_{m1} - t_p}{\tau}} - 2e^{-\frac{t_{m1} - t_p + t_{m2} - t_p}{\tau}} + e^{-2\frac{t_{m2} - t_p}{\tau}} \right) dt_p + \int_{t_p=t_{m1}}^{t_p=t_{m2}} e^{-2\frac{t_{m2} - t_p}{\tau}} dt_p \right) \\ \sigma_{sum}^2 &= \frac{\pi Q_i}{\tau^2} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( \left[ \frac{\tau}{2} e^{-2\frac{t_{m1} - t_p}{\tau}} - \tau e^{-\frac{t_{m1} - t_p + t_{m2} - t_p}{\tau}} + \frac{\tau}{2} e^{-2\frac{t_{m2} - t_p}{\tau}} \right]_{t_0}^{t_{m1}} + \left[ \frac{\tau}{2} e^{-2\frac{t_{m2} - t_p}{\tau}} \right]_{t_{m1}}^{t_{m2}} \right) \\ \sigma_{sum}^2 &= \frac{\pi Q_i}{\tau^2} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( \frac{\tau}{2} - \tau e^{-\frac{t_{12}}{\tau}} + \frac{\tau}{2} e^{-2\frac{t_{12}}{\tau}} - \frac{\tau}{2} e^{-2\frac{t_{01}}{\tau}} + \tau e^{-\frac{t_{01} + t_{02}}{\tau}} - \frac{\tau}{2} e^{-2\frac{t_{02}}{\tau}} \right) \\ &\quad + \frac{\pi Q_i}{\tau^2} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( \frac{\tau}{2} - \frac{\tau}{2} e^{-2\frac{t_{12}}{\tau}} \right) \\ \sigma_{sum}^2(t_\infty) &= \frac{\pi Q_i}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( 1 - e^{-\frac{t_{12}}{\tau}} \right) \\ \sigma_{sum}^2(t_\infty) &= \frac{2kTr}{\tau} \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \left( 1 - e^{-\frac{t_{12}}{\tau}} \right) \end{aligned}$$

B. Thermal noise of a charge amp and RC filter followed by a Correlated Double Sampler in the frequency domain.

$$\sigma_{sum}^2(t_\infty) = \int_0^\infty Q_i |H_r(jw)|^2 dw$$

$$H_r(s) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{1}{1 + s\tau} (1 - e^{-st_{12}})$$

$$|H_r(jw)|^2 = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{2(1 - \cos wt_{12})}{(1 + w^2 \tau^2)}$$

$$\sigma_{sum}^2(t_\infty) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^\infty \frac{2(1 - \cos wt_{12})}{1 + (w\tau)^2} dw$$

$$\sigma_{sum}^2(t_\infty) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( \left[ \frac{2}{\tau} \tan^{-1} w\tau \right]_0^\infty - \frac{\pi}{\tau} e^{-\frac{t_{12}}{\tau}} \right)$$

$$\sigma_{sum}^2(t_\infty) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{\pi}{\tau} \left( 1 - e^{-\frac{t_{12}}{\tau}} \right)$$

$$\sigma_{sum}^2(t_\infty) = Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \left( 1 - e^{-\frac{t_{12}}{\tau}} \right)$$

$$\sigma_{sum}^2(t_\infty) = \frac{2kTr}{\tau} \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \left( 1 - e^{-\frac{t_{12}}{\tau}} \right)$$

B. Shot noise through charge amp, RC filter and a

Correlated Double Sampler in the time domain.

$$\sigma_{sum}^2 = \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_s \left( -h(t_{m1} - t_p) + h(t_{m2} - t_p) \right)^2 dt_p + \int_{t_p=t_{m1}}^{t_p=t_{m2}} \pi Q_s \left( h(t_{m2} - t_p) \right)^2 dt_p$$

$$h_i(t_m - t_p) = \frac{A}{c_{tot} + A c_f} \left( 1 - e^{-\frac{t_m - t_p}{\tau}} \right)$$

$$\sigma_{sum}^2 = \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_s \left( \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( e^{-\frac{t_{m1} - t_p}{\tau}} - e^{-\frac{t_{m2} - t_p}{\tau}} \right) \right)^2 dt_p$$

$$+ \int_{t_p=t_{m1}}^{t_p=t_{m2}} \pi Q_s \left( \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( 1 - e^{-\frac{t_{m2} - t_p}{\tau}} \right) \right)^2 dt_p$$

$$\sigma_{sum}^2 = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \int_{t_p=t_0}^{t_p=t_{m1}} \left( e^{-2\frac{t_{m1} - t_p}{\tau}} - 2e^{-\frac{t_{m1} - t_p + t_{m2} - t_p}{\tau}} + e^{-2\frac{t_{m2} - t_p}{\tau}} \right) dt_p$$

$$+ Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \int_{t_p=t_{m1}}^{t_p=t_{m2}} \left( 1 - 2e^{-\frac{t_{m2} - t_p}{\tau}} + e^{-2\frac{t_{m2} - t_p}{\tau}} \right) dt_p$$

$$\sigma_{sum}^2 = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left[ \frac{\tau}{2} e^{-2\frac{t_{m1} - t_p}{\tau}} - \tau e^{-\frac{t_{m1} - t_p + t_{m2} - t_p}{\tau}} + \frac{\tau}{2} e^{-2\frac{t_{m2} - t_p}{\tau}} \right]_{t_0}^{t_{m1}}$$

$$+ Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left[ t_p - 2\tau e^{-\frac{t_{m2} - t_p}{\tau}} + \frac{\tau}{2} e^{-2\frac{t_{m2} - t_p}{\tau}} \right]_{t_{m1}}^{t_{m2}}$$

$$\sigma_{sum}^2 = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( \frac{\tau}{2} - \tau e^{-\frac{t_{12}}{\tau}} + \frac{\tau}{2} e^{-2\frac{t_{12}}{\tau}} - \frac{\tau}{2} e^{-2\frac{t_{01}}{\tau}} + \tau e^{-\frac{t_{01} + t_{02}}{\tau}} - \frac{\tau}{2} e^{-2\frac{t_{02}}{\tau}} \right)$$

$$+ Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( t_{12} - 2\tau + \frac{\tau}{2} + 2\tau e^{-\frac{t_{12}}{\tau}} - \frac{\tau}{2} e^{-2\frac{t_{12}}{\tau}} \right)$$

$$\sigma_{sum}^2(t_\infty) = \pi Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( t_{12} - \tau + \tau e^{-\frac{t_{12}}{\tau}} \right)$$

$$\sigma_{sum}^2(t_\infty) = iq \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( t_{12} - \tau + \tau e^{-\frac{t_{12}}{\tau}} \right)$$

B. Shot noise of a charge amplifier and RC filter and a Correlated Double Sampler in the frequency domain.

$$\sigma_{sum}^2(t_\infty) = \int_0^\infty Q_s |H_i(j\omega)|^2 d\omega$$

$$H_i(s) = \left( \frac{A}{c_{tot} + A c_f} \right) \frac{1}{s(1 + s\tau)} (1 - e^{-st_{12}})$$

$$\sigma_{sum}^2(t_\infty) = \int_0^\infty Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{2(1 - \cos \omega t_{12})}{\omega^2 (1 + (\omega\tau)^2)} d\omega$$

$$\sigma_{sum}^2(t_\infty) = 2\tau^2 Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \int_0^\infty \left( \frac{1 - \cos \omega t_{12}}{(\omega\tau)^2} - \frac{1}{1 + (\omega\tau)^2} + \frac{\cos \omega t_{12}}{1 + (\omega\tau)^2} \right) d\omega$$

$$\sigma_{sum}^2(t_\infty) = 2\tau^2 Q_s \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( \frac{\pi t_{12}}{2\tau^2} - \frac{\pi}{2\tau} + \frac{\pi}{2\tau} e^{-\frac{t_{12}}{\tau}} \right)$$

$$\sigma_{sum}^2(t_\infty) = iq \left( \frac{A}{c_{tot} + A c_f} \right)^2 \left( t_{12} - \tau + \tau e^{-\frac{t_{12}}{\tau}} \right)$$

B. Flicker noise of charge amplifier with an CR filter followed by a Correlated Double Sampler in the frequency domain.

$$\sigma_{sum}^2(t_\infty) = \int_0^\infty \frac{Q_f}{w} |H_r(jw)|^2 dw$$

$$H_r(s) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \frac{1}{1 + s\tau} (1 - e^{-st_{12}})$$

$$|H_r(jw)|^2 = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{2(1 - \cos wt_{12})}{(1 + w^2\tau^2)}$$

$$\sigma_{sum}^2(t_\infty) = \int_0^\infty \frac{Q_f}{w} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{2(1 - \cos wt_{12})}{(1 + w^2\tau^2)} dw$$

$$= Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^\infty \frac{2(1 - \cos wt_{12})}{w(1 + (w\tau)^2)} dw$$

Substituting  $x = w\tau$

$$\sigma_{sum}^2(t_\infty) = Q_f \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^\infty \frac{2 \left( 1 - \cos x \frac{t_{12}}{\tau} \right)}{x(1 + x^2)} dx$$

$$\sigma_{sum}^2(t_\infty) = \frac{Kf_2}{WLCox} \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \text{(Factor1)}$$

and

$$ENC^2 = Q_f c_{tot}^2 \frac{1}{\left( 1 - e^{-\frac{t_{12}}{\tau}} \right)^2} \int_0^\infty \frac{2(1 - \cos wt_{12})}{w(1 + (w\tau)^2)} dw$$

$$ENC^2 = \frac{Kf_2}{WLC_{ox}} c_{tot}^2 \text{(Factor2)}$$

B. Flicker noise of charge amplifier with an RC filter followed by a Correlated Double Sampler in the frequency domain (cont).

Factor 1 and Factor 2 can be tabulated for different  $\left(\frac{t_{12}}{\tau}\right)$  values:-

$\left(\frac{t_{12}}{\tau}\right)$	0.5	1	2	3	4	6	8	10
Factor 1	0.42	1.05	2.23	3.12	3.77	4.67	5.28	5.74
Factor 2	2.68	2.64	2.98	3.45	3.92	4.69	5.28	5.74

$\left(\frac{t_{12}}{\tau}\right)$	12	16	32	50	100	1000	10000
Factor 1	6.11	6.69	8.08	8.98	10.4	15.0	20.1
Factor 2	6.11	6.69	8.08	8.98	10.4	15.0	20.1

Best ENC at  $\frac{t_{12}}{\tau} = 0.8$  when Factor 2 = 2.62

B. Thermal noise of charge amp and CR-RC filter with triple sampled deconvolution.

$$\sigma_{sum}^2 = \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_i \left( b_1 h_r(t_{m1} - t_p) + b_2 h_r(t_{m2} - t_p) + b_3 h_r(t_{m3} - t_p) \right)^2 dt$$

$$+ \int_{t_p=t_{m1}}^{t_p=t_{m2}} \pi Q_i \left( b_2 h_r(t_{m2} - t_p) + b_3 h_r(t_{m3} - t_p) \right)^2 dt_p + \int_{t_p=t_{m2}}^{t_p=t_{m3}} \pi Q_i \left( b_3 h_r(t_{m3} - t_p) \right)^2 dt_p$$

$$h_r(t_{m1} - t_p) = \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \left( \frac{1}{\tau} - \frac{(t_{m1} - t_p)}{(\tau)^2} \right) e^{-\frac{(t_{m1} - t_p)}{\tau}}$$

$$h_r(t_{m1} - t_p) = \frac{-1}{(\tau)^2} \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \left( (t_{m1} - t_p) - \tau \right) e^{-\frac{(t_{m1} - t_p)}{\tau}}$$

and  $t_{m3} - t_{m2} = t_{m2} - t_{m1} = \Delta$  the multiple sample formula gives

$$\sigma_{sum}^2 = \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_i \left( b_1 h_r(t_{m1} - t_p) + b_2 h_r(t_{m1} + \Delta - t_p) + b_3 h_r(t_{m1} + 2\Delta - t_p) \right)^2 dt_p$$

$$+ \int_{t_p=t_{m1}}^{t_p=t_{m1} + \Delta} \pi Q_i \left( b_2 h_r(t_{m1} + \Delta - t_p) + b_3 h_r(t_{m1} + 2\Delta - t_p) \right)^2 dt_p$$

$$+ \int_{t_p=t_{m1} + \Delta}^{t_p=t_{m1} + 2\Delta} \pi Q_i \left( b_3 h_r(t_{m1} + 2\Delta - t_p) \right)^2 dt_p$$

substituting  $-t_p' = t_{m1} - t_p$  or  $t_p' = -t_{m1} + t_p$  or  $-t_p = -t_{m1} - t_p'$

$$\sigma_{sum}^2 = \int_{t_p=t_{01}}^{t_p=0} \pi Q_i \left( b_1 h_r(-t_p') + b_2 h_r(\Delta - t_p') + b_3 h_r(2\Delta - t_p') \right)^2 dt_p'$$

$$+ \int_{t_p=0}^{t_p=\Delta} \pi Q_i \left( b_2 h_r(\Delta - t_p') + b_3 h_r(2\Delta - t_p') \right)^2 dt_p'$$

$$+ \int_{t_p=\Delta}^{t_p=2\Delta} \pi Q_i \left( b_3 h_r(2\Delta - t_p') \right)^2 dt_p'$$

$$\frac{\sigma_{sum}^2}{\pi Q_i \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2} = \int_{t_p=t_{01}}^{t_p=0} \left( b_1 (-t_p' - \tau) e^{-\frac{(-t_p')}{\tau}} + b_2 (\Delta - t_p' - \tau) e^{-\frac{(\Delta - t_p')}{\tau}} + b_3 (2\Delta - t_p' - \tau) e^{-\frac{(2\Delta - t_p')}{\tau}} \right)^2 dt_p'$$

$$+ \int_{t_p=0}^{t_p=\Delta} \left( b_2 (\Delta - t_p' - \tau) e^{-\frac{(\Delta - t_p')}{\tau}} + b_3 (2\Delta - t_p' - \tau) e^{-\frac{(2\Delta - t_p')}{\tau}} \right)^2 dt_p'$$

B. Thermal noise of charge amp and CR-RC filter with triple sampled deconvolution (cont).

$$\begin{aligned}
 & + \int_{t_p'=\Delta}^{t_p'=2\Delta} \left( b_3 (2\Delta - t_p' - \tau) e^{-\frac{(2\Delta - t_p')}{\tau}} \right)^2 dt_p', \\
 \text{if } b_1 &= \frac{\tau}{\Delta} e^{-\left(\frac{\tau+\Delta}{\tau}\right)}, b_2 = -\frac{2\tau}{\Delta} e^{-\left(\frac{\tau}{\tau}\right)}, b_3 = \frac{\tau}{\Delta} e^{-\left(\frac{\tau-\Delta}{\tau}\right)} \\
 & \frac{\sigma_{sum}^2}{\frac{\pi Q_i}{\tau^4} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2} \\
 & = \int_{t_p'=t_{01}}^{t_p'=0} \left( \frac{\tau}{\Delta} e^{-\left(\frac{\tau+\Delta}{\tau}\right)} (-t_p' - \tau) e^{-\frac{(-t_p')}{\tau}} - \frac{2\tau}{\Delta} e^{-\left(\frac{\tau}{\tau}\right)} (\Delta - t_p' - \tau) e^{-\frac{(\Delta - t_p')}{\tau}} + \frac{\tau}{\Delta} e^{-\left(\frac{\tau-\Delta}{\tau}\right)} (2\Delta - t_p' - \tau) e^{-\frac{(2\Delta - t_p')}{\tau}} \right)^2 dt_p', \\
 & + \int_{t_p'=0}^{t_p'=\Delta} \left( -\frac{2\tau}{\Delta} e^{-\left(\frac{\tau}{\tau}\right)} (\Delta - t_p' - \tau) e^{-\frac{(\Delta - t_p')}{\tau}} + \frac{\tau}{\Delta} e^{-\left(\frac{\tau-\Delta}{\tau}\right)} (2\Delta - t_p' - \tau) e^{-\frac{(2\Delta - t_p')}{\tau}} \right)^2 dt_p', \\
 & + \int_{t_p'=\Delta}^{t_p'=2\Delta} \left( \frac{\tau}{\Delta} e^{-\left(\frac{\tau-\Delta}{\tau}\right)} (2\Delta - t_p' - \tau) e^{-\frac{(2\Delta - t_p')}{\tau}} \right)^2 dt_p', \\
 & \frac{\sigma_{sum}^2}{\frac{\pi Q_i}{\tau^2 \Delta^2} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2} = \int_{t_p'=t_{01}}^{t_p'=0} \left( (-t_p' - \tau) - 2(\Delta - t_p' - \tau) + (2\Delta - t_p' - \tau) \right)^2 e^{-2\frac{(\tau+\Delta - t_p')}{\tau}} dt_p', \\
 & + \int_{t_p'=0}^{t_p'=\Delta} \left( -2(\Delta - t_p' - \tau) + (2\Delta - t_p' - \tau) \right)^2 e^{-2\frac{(\tau+\Delta - t_p')}{\tau}} dt_p', \\
 & + \int_{t_p'=\Delta}^{t_p'=2\Delta} \left( 2\Delta - t_p' - \tau \right)^2 e^{-2\frac{(\tau+\Delta - t_p')}{\tau}} dt_p',
 \end{aligned}$$



B. Thermal noise of charge amp and CR-RC filter with triple sampled deconvolution (cont).

$$\begin{aligned}
 \frac{\sigma_{sum}^2}{\tau^2 \Delta^2} &= 0 + \int_{t_p'=0}^{t_p'=\Delta} (t_p'+\tau)^2 e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} dt_p', \\
 \frac{\pi Q_t}{\tau^2 \Delta^2} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 &+ \int_{t_p'=\Delta}^{t_p'=2\Delta} \left( 4\Delta^2 - 2(t_p'+\tau)2\Delta + (t_p'+\tau)^2 \right) e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} dt_p', \\
 \frac{\sigma_{sum}^2}{\tau^2 \Delta^2} &= \int_{t_p'=0}^{t_p'=2\Delta} (t_p'+\tau)^2 e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} dt_p', \\
 \frac{\pi Q_t}{\tau^2 \Delta^2} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 &+ \int_{t_p'=\Delta}^{t_p'=2\Delta} \left( 4\Delta^2 - 2(t_p'+\tau)2\Delta \right) e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} dt_p', \\
 \frac{\sigma_{sum}^2}{\tau^2 \Delta^2} &= \int_{t_p'=0}^{t_p'=2\Delta} \left( (t_p')^2 + 2t_p'\tau + (\tau)^2 \right) e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} dt_p', \\
 \frac{\pi Q_t}{\tau^2 \Delta^2} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 &+ \int_{t_p'=\Delta}^{t_p'=2\Delta} 4\Delta(-t_p'-\tau+\Delta) e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} dt_p', \\
 \frac{\sigma_{sum}^2}{\tau^2 \Delta^2} &= \left[ \left( \frac{(t_p')^2}{2} \tau - \frac{2t_p'\tau^2}{4} + \frac{2\tau^3}{8} + \frac{2t_p'\tau^2}{2} - \frac{2\tau^3}{4} + \frac{\tau^3}{2} \right) e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} \right]_0^{2\Delta} \\
 &- \left[ 4\Delta \left( \frac{t_p'\tau}{2} - \frac{\tau^2}{4} + \frac{(\tau-\Delta)\tau}{2} \right) e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} \right]_{\Delta}^{2\Delta} \\
 \frac{\sigma_{sum}^2}{\tau^2 \Delta^2} &= \left[ \left( \frac{(t_p')^2}{2} \tau + \frac{2t_p'\tau^2}{4} + \frac{2\tau^3}{8} \right) e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} \right]_0^{2\Delta} \\
 \frac{\pi Q_t}{\tau^2 \Delta^2} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 &- \left[ 4\Delta \left( \frac{t_p'\tau}{2} - \frac{\tau^2}{4} + \frac{(\tau^2-\Delta)\tau}{2} \right) e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} \right]_{\Delta}^{2\Delta}
 \end{aligned}$$

B. Thermal noise of charge amp and CR-RC filter with triple sampled deconvolution (cont).

$$\frac{\sigma_{sum}^2}{\left(\frac{\pi Q_i}{r\tau^2\Delta^2} \left(\frac{A c_{tot}}{gm(c_{tot} + A c_f)}\right)\right)^2} = \left(\frac{(2\Delta)^2\tau}{2} + \frac{22\Delta\tau}{4} + \frac{2\tau^3}{8}\right) e^{-2\frac{(\tau+\Delta-2\Delta)}{\tau}} - \left(\frac{2\tau^3}{8}\right) e^{-2\frac{(\tau+\Delta)}{\tau}}$$

$$- 4\Delta\left(\frac{2\Delta\tau}{2} + \frac{\tau^2}{4} - \frac{\Delta\tau}{2}\right) e^{-2\frac{(\tau+\Delta-2\Delta)}{\tau}} + 4\Delta\left(\frac{\Delta\tau}{2} + \frac{\tau^2}{4} - \frac{\Delta\tau}{2}\right) e^{-2\frac{(\tau)}{\tau}}$$

$$\frac{\sigma_{sum}^2}{\left(\frac{\pi Q_i}{r\tau^2\Delta^2} \left(\frac{A c_{tot}}{gm(c_{tot} + A c_f)}\right)\right)^2} = \left(\frac{(2\Delta)^2\tau}{2} + \frac{4\Delta\tau^2}{4} + \frac{2\tau^3}{8}\right) e^{-2\frac{(\tau-\Delta)}{\tau}} - \left(\frac{2\tau^3}{8}\right) e^{-2\frac{(\tau+\Delta)}{\tau}}$$

$$- \left(\frac{4\Delta^2\tau}{2} + \frac{4\Delta\tau^2}{4}\right) e^{-2\frac{(\tau-\Delta)}{\tau}} + 4\Delta\left(\frac{\tau^2}{4}\right) e^{-2\frac{(\tau)}{\tau}}$$

$$\frac{\sigma_{sum}^2}{\left(\frac{\pi Q_i}{r\tau^2\Delta^2} \left(\frac{A c_{tot}}{gm(c_{tot} + A c_f)}\right)\right)^2} = \left(\frac{2\tau^3}{8}\right) e^{-2\frac{(\tau-\Delta)}{\tau}} - \left(\frac{2\tau^3}{8}\right) e^{-2\frac{\tau+\Delta}{\tau}} + 4\Delta\left(\frac{\tau}{4}\right) e^{-2\frac{(\tau)}{\tau}}$$

$$\sigma_{sum}^2 = \pi Q_i \left(\frac{A c_{tot}}{gm(c_{tot} + A c_f)}\right)^2 \frac{1}{4\Delta} \left( \left(\frac{\tau}{\Delta}\right) e^{-2\frac{(\tau-\Delta)}{\tau}} + 4e^{-2} - \left(\frac{\tau}{\Delta}\right) e^{-2\frac{\tau+\Delta}{\tau}} \right)$$

$$\sigma_{sum}^2 = \pi Q_i \left(\frac{A c_{tot}}{gm(c_{tot} + A c_f)}\right)^2 \frac{1}{4\Delta} \left( -\left(\frac{\tau}{\Delta}\right) e^{-2\frac{(\tau+\Delta)}{\tau}} + 4e^{-2} + \left(\frac{\tau}{\Delta}\right) e^{-2\frac{\tau-\Delta}{\tau}} \right)$$

if  $\tau = 2\Delta$

$$\sigma_{sum}^2 = 2kTr \left(\frac{A c_{tot}}{(c_f + A c_{tot})}\right)^2 \frac{1}{2\tau} (-2e^{-3} + 4e^{-2} + 2e^{-1})$$

B. Thermal noise of charge amp and CR-RC filter with triple sampled in the frequency domain.

$$o_{\text{sum}}^2(t_{\infty}) = \int_0^{\infty} Q_t |H_r(jw)|^2 dw$$

if  $\tau = 2\Delta$

$$|H_r(jw)|^2 = \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \frac{w^2 \tau^2}{(1 + w^2 \tau^2)^2} \frac{4}{e^2} \left( e^{\frac{1}{2}} + e^{-\frac{1}{2}} - 2 \cos \frac{(w\tau)}{2} \right)^2$$

$$o_{\text{sum}}^2(t_{\infty}) = Q_t \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^{\infty} \frac{w^2 \tau^2}{(1 + w^2 \tau^2)^2} \frac{4}{e^2} \left( e^{\frac{1}{2}} + e^{-\frac{1}{2}} - 2 \cos \frac{(w\tau)}{2} \right)^2 dw$$

$$o_{\text{sum}}^2(t_{\infty}) = \frac{Q_t}{\tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^{\infty} \frac{w^2 \tau^2}{(1 + w^2 \tau^2)^2} \frac{4}{e^2} \left( e^{\frac{1}{2}} + e^{-\frac{1}{2}} - 2 \cos \frac{(w\tau)}{2} \right)^2 dw \tau$$

$$o_{\text{sum}}^2(t_{\infty}) = \frac{4Q_t}{e^2 \tau} \left( \frac{A c_{tot}}{gm(c_{tot} + A c_f)} \right)^2 \int_0^{\infty} \frac{x^2}{(1 + x^2)^2} \left( e^{\frac{1}{2}} + e^{-\frac{1}{2}} - 2 \cos \frac{x}{2} \right)^2 dx$$

$$o_{\text{sum}}^2(t_{\infty}) = \left( \frac{A c_{tot}}{(c_{tot} + A c_f)} \right)^2 \frac{8.7 kTr}{e^2 \tau}$$

B. Shot noise of charge amp and CR-RC filter with triple sampled deconvolution.

$$\begin{aligned}\sigma_{sum}^2 &= \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_s \left( b_1 h_i(t_{m1} - t_p) + b_2 h_i(t_{m2} - t_p) + b_3 h_i(t_{m3} - t_p) \right)^2 dt_p \\ &+ \int_{t_p=t_{m1}}^{t_p=t_{m2}} \pi Q_s \left( b_2 h_i(t_{m2} - t_p) + b_3 h_i(t_{m3} - t_p) \right)^2 dt_p + \int_{t_p=t_{m2}}^{t_p=t_{m3}} \pi Q_s \left( b_3 h_i(t_{m3} - t_p) \right)^2 dt_p\end{aligned}$$

$$h_i(t_{m1} - t_p) = \frac{A}{c_{tot} + A c_f} \frac{(t_{m1} - t_p)}{\tau} e^{-\frac{(t_{m1} - t_p)}{\tau}}$$

with  $t_{m3} - t_{m2} = t_{m2} - t_{m1} = t_{m1} - t_0 = \Delta$  the multiple sample formula gives

$$\begin{aligned}\sigma_{sum}^2 &= \int_{t_p=t_0}^{t_p=t_{m1}} \pi Q_s \left( b_1 h_i(t_{m1} - t_p) + b_2 h_i(t_{m1} + \Delta - t_p) + b_3 h_i(t_{m1} + 2\Delta - t_p) \right)^2 dt_p \\ &+ \int_{t_p=t_{m1}}^{t_p=t_{m1} + \Delta} \pi Q_s \left( b_2 h_i(t_{m1} + \Delta - t_p) + b_3 h_i(t_{m1} + 2\Delta - t_p) \right)^2 dt_p \\ &+ \int_{t_p=t_{m1} + \Delta}^{t_p=t_{m1} + 2\Delta} \pi Q_s \left( b_3 h_i(t_{m1} + 2\Delta - t_p) \right)^2 dt_p\end{aligned}$$

substituting  $-t_p' = t_{m1} - t_p$  or  $t_p' = t_{m1} + t_p$  or  $-t_p' = -t_{m1} - t_p$ ,

$$\begin{aligned}\sigma_{sum}^2 &= \int_{t_p'=t_{01}}^{t_p'=0} \pi Q_s \left( b_1 h_i(-t_p') + b_2 h_i(\Delta - t_p') + b_3 h_i(2\Delta - t_p') \right)^2 dt_p' \\ &+ \int_{t_p'=0}^{t_p'=\Delta} \pi Q_s \left( b_2 h_i(\Delta - t_p') + b_3 h_i(2\Delta - t_p') \right)^2 dt_p' \\ &+ \int_{t_p'=\Delta}^{t_p'=2\Delta} \pi Q_s \left( b_3 h_i(2\Delta - t_p') \right)^2 dt_p'\end{aligned}$$

$$\begin{aligned}\frac{\sigma_{sum}^2}{\left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{\pi Q_s}{\tau^2}} &= \int_{t_p'=t_{01}}^{t_p'=0} \left( b_1 (-t_p') e^{-\frac{(-t_p')}{\tau}} + b_2 (\Delta - t_p') e^{-\frac{(\Delta - t_p')}{\tau}} + b_3 (2\Delta - t_p') e^{-\frac{(2\Delta - t_p')}{\tau}} \right)^2 dt_p' \\ &+ \int_{t_p'=0}^{t_p'=\Delta} \left( b_2 (\Delta - t_p') e^{-\frac{(\Delta - t_p')}{\tau}} + b_3 (2\Delta - t_p') e^{-\frac{(2\Delta - t_p')}{\tau}} \right)^2 dt_p' \\ &+ \int_{t_p'=\Delta}^{t_p'=2\Delta} \left( b_3 (2\Delta - t_p') e^{-\frac{(2\Delta - t_p')}{\tau}} \right)^2 dt_p'\end{aligned}$$

B. Shot noise of charge amp and CR-RC filter with triple sampled deconvolution (cont).

$$\text{if } b_1 = \frac{\tau}{\Delta} e^{-\left(\frac{\tau+\Delta}{\tau}\right)}, b_2 = -\frac{2\tau}{\Delta} e^{-\left(\frac{\tau}{\tau}\right)}, b_3 = \frac{\tau}{\Delta} e^{-\left(\frac{\tau-\Delta}{\tau}\right)}$$

$$\begin{aligned} \frac{\sigma_{sum}^2}{\left(\frac{A}{c_{tot} + A c_f}\right)^2} &= \frac{\pi Q_s}{\tau^2} \\ &\int_{t_p'=-t_{01}}^{t_p'=0} \left( \frac{\tau}{\Delta} e^{-\left(\frac{\tau+\Delta}{\tau}\right)} (-t_p') e^{-\left(\frac{-t_p'}{\tau}\right)} - \frac{2\tau}{\Delta} e^{-\left(\frac{\tau}{\tau}\right)} (\Delta - t_p') e^{-\left(\frac{\Delta-t_p'}{\tau}\right)} + \frac{\tau}{\Delta} e^{-\left(\frac{\tau-\Delta}{\tau}\right)} (2\Delta - t_p') e^{-\left(\frac{2\Delta-t_p'}{\tau}\right)} \right)^2 dt_p' \\ &+ \int_{t_p'=0}^{t_p'=\Delta} \left( -\frac{2\tau}{\Delta} e^{-\left(\frac{\tau}{\tau}\right)} (\Delta - t_p') e^{-\left(\frac{\Delta-t_p'}{\tau}\right)} + \frac{\tau}{\Delta} e^{-\left(\frac{\tau-\Delta}{\tau}\right)} (2\Delta - t_p') e^{-\left(\frac{2\Delta-t_p'}{\tau}\right)} \right)^2 dt_p' \\ &+ \int_{t_p'=\Delta}^{t_p'=2\Delta} \left( \frac{\tau}{\Delta} e^{-\left(\frac{\tau-\Delta}{\tau}\right)} (2\Delta - t_p') e^{-\left(\frac{2\Delta-t_p'}{\tau}\right)} \right)^2 dt_p' \\ \frac{\sigma_{sum}^2}{\left(\frac{A}{c_{tot} + A c_f}\right)^2} &= \int_{t_p'=t_{01}}^{t_p'=0} \left( (-t_p') - 2(\Delta - t_p') + (2\Delta - t_p') \right)^2 e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} dt_p' \\ &+ \int_{t_p'=0}^{t_p'=\Delta} \left( -2(\Delta - t_p') + (2\Delta - t_p') \right)^2 e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} dt_p' \\ &+ \int_{t_p'=\Delta}^{t_p'=2\Delta} (2\Delta - t_p')^2 e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} dt_p' \\ \frac{\sigma_{sum}^2}{\left(\frac{A}{c_{tot} + A c_f}\right)^2} &= 0 + \int_{t_p'=0}^{t_p'=\Delta} (t_p')^2 e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} dt_p' \\ &+ \int_{t_p'=\Delta}^{t_p'=2\Delta} (4\Delta^2 - 2t_p' 2\Delta + (t_p')^2) e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} dt_p' \\ \frac{\sigma_{sum}^2}{\left(\frac{A}{c_{tot} + A c_f}\right)^2} &= \left[ \left( \frac{(t_p')^2 \tau}{2} - \frac{t_p' \tau^2}{2} + \frac{\tau^3}{4} \right) e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} \right]_0^\Delta \end{aligned}$$

B. Shot noise of charge amp and CR-RC filter with triple sampled deconvolution (cont).

$$\begin{aligned}
 & + \left[ \left( \frac{4\Delta^2\tau}{2} - \frac{2t_p'\tau 2\Delta}{2} + \frac{2\tau^2 2\Delta}{4} + \frac{(t_p')^2\tau}{2} - \frac{t_p'\tau^2}{2} + \frac{\tau^3}{4} \right) e^{-2\frac{(\tau+\Delta-t_p')}{\tau}} \right]_{\Delta}^{2\Delta} \\
 \frac{\sigma_{sum}^2}{\left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{\pi Q_s}{\Delta^2}} & = \left( \frac{\Delta^2\tau}{2} - \frac{\Delta\tau^2}{2} + \frac{\tau^3}{4} \right) e^{-2} - \left( \frac{\tau^3}{4} \right) e^{-2\frac{(\tau+\Delta)}{\tau}} \\
 & + \left( \frac{4\Delta^2\tau}{2} - \frac{4\Delta\tau 2\Delta}{2} + \frac{2\tau^2\Delta}{4} + \frac{4\Delta^2\tau}{2} - \frac{2\Delta\tau^2}{2} + \frac{\tau^3}{4} \right) e^{-2\frac{(\tau-\Delta)}{\tau}} \\
 & - \left( \frac{4\Delta^2\tau}{2} - \frac{2\Delta\tau 2\Delta}{2} + \frac{2\tau^2 2\Delta}{4} + \frac{\Delta^2\tau}{2} - \frac{\Delta\tau^2}{2} + \frac{\tau^3}{4} \right) e^{-2} \\
 \frac{\sigma_{sum}^2}{\left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{\pi Q_s}{\Delta^2}} & = - \left( \frac{\tau^3}{4} \right) e^{-2\frac{(\tau+\Delta)}{\tau}} + \left( \frac{\tau^3}{4} \right) e^{-2\frac{(\tau-\Delta)}{\tau}} - (\tau^2\Delta) e^{-2} \\
 \sigma_{sum}^2 & = \left( \frac{A}{c_{tot} + A c_f} \right)^2 \pi Q_s \frac{\tau^2}{4\Delta} \left( - \left( \frac{\tau}{\Delta} \right) e^{-2\frac{(\tau+\Delta)}{\tau}} - 4e^{-2} + \left( \frac{\tau}{\Delta} \right) e^{-2\frac{(\tau-\Delta)}{\tau}} \right) \\
 \text{if } \tau & = 2\Delta \\
 \sigma_{sum}^2 & = \left( \frac{A}{c_{tot} + A c_f} \right)^2 iq \frac{2\tau}{4} (-2e^{-3} - 4e^{-2} + 2e^{-1}) \\
 \sigma_{sum}^2 & = \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{iq}{e^2} \tau (-e^{-1} - 2 + e^1) \\
 \sigma_{sum}^2 & = \left( \frac{A}{c_{tot} + A c_f} \right)^2 \frac{iq}{e^2} \tau \cdot 0.35
 \end{aligned}$$

B. Shot noise of charge amp and CR-RC filter with triple sampled deconvolution

$$o_{\text{sum}}^2(t_\infty) = \int_0^\infty Q_s |H_i(jw)|^2 dw$$

if  $\tau = 2\Delta$

$$|H_i(jw)|^2 = \left( \frac{A}{(c_{\text{tot}} + A c_f)} \right)^2 \frac{\tau^2}{(1 + w^2 \tau^2)^2} \frac{4}{e^2} \left( e^{\frac{1}{2}} + e^{-\frac{1}{2}} - 2 \cos \frac{(w\tau)}{2} \right)^2$$

$$o_{\text{sum}}^2(t_\infty) = Q_s \left( \frac{A}{(c_{\text{tot}} + A c_f)} \right)^2 \int_0^\infty \frac{\tau^2}{(1 + w^2 \tau^2)^2} \frac{4}{e^2} \left( e^{\frac{1}{2}} + e^{-\frac{1}{2}} - 2 \cos \frac{(w\tau)}{2} \right)^2 dw$$

$$o_{\text{sum}}^2(t_\infty) = Q_s \left( \frac{A}{(c_{\text{tot}} + A c_f)} \right)^2 \int_0^\infty \frac{\tau}{(1 + w^2 \tau^2)^2} \frac{4}{e^2} \left( e^{\frac{1}{2}} + e^{-\frac{1}{2}} - 2 \cos \frac{(w\tau)}{2} \right)^2 dw \tau$$

$$o_{\text{sum}}^2(t_\infty) = \frac{4Q_s \tau}{e^2} \left( \frac{A}{(c_{\text{tot}} + A c_f)} \right)^2 \int_0^\infty \frac{1}{(1 + x^2)^2} \left( e^{\frac{1}{2}} + e^{-\frac{1}{2}} - 2 \cos \frac{x}{2} \right)^2 dx$$

$$o_{\text{sum}}^2(t_\infty) = \left( \frac{A}{(c_{\text{tot}} + A c_f)} \right)^2 \frac{0.35}{e^2} iq \tau$$

B. Flicker noise of charge amp and CR-RC filter with triple sampled deconvolution.

$$\sigma_{sum}^2(t_\infty) = \int_0^\infty \frac{Q_f}{w} |H_r(jw)|^2 dw$$

$$b_1 = \frac{\tau}{\Delta} e^{-\left(\frac{\tau+\Delta}{\tau}\right)}, b_2 = -\frac{2\tau}{\Delta} e^{-\left(\frac{\tau}{\tau}\right)}, b_3 = \frac{\tau}{\Delta} e^{-\left(\frac{\tau-\Delta}{\tau}\right)}$$

$$\frac{H_r(s)}{\left(\frac{A c_{tot}}{gm(c_{tot} + A c_f)}\right)} = \frac{s\tau}{(1+s\tau)^2} \left( \frac{\tau}{\Delta} e^{-\left(\frac{\tau+\Delta}{\tau}\right)} e^{s\Delta} - \frac{2\tau}{\Delta} e^{-\left(\frac{\tau}{\tau}\right)} + \frac{\tau}{\Delta} e^{-\left(\frac{\tau-\Delta}{\tau}\right)} e^{-s\Delta} \right)$$

$$\frac{|H_r(jw)|^2}{\left(\frac{A c_{tot}}{gm(c_{tot} + A c_f)}\right)^2} = \frac{w^2 \tau^2}{(1+w^2 \tau^2)^2} \left(\frac{\tau}{\Delta}\right)^2 e^{-2} \left( e^{-\left(\frac{\Delta}{\tau}\right)} e^{jw\Delta} - 2 + e^{-\left(\frac{-\Delta}{\tau}\right)} e^{-jw\Delta} \right) \left( e^{-\left(\frac{\Delta}{\tau}\right)} e^{-jw\Delta} - 2 + e^{-\left(\frac{-\Delta}{\tau}\right)} e^{jw\Delta} \right)$$

$$= \frac{w^2 \tau^2}{(1+w^2 \tau^2)^2} \left(\frac{\tau}{\Delta e}\right)^2$$

$$\left( e^{-\left(\frac{2\Delta}{\tau}\right)} - 2e^{-\left(\frac{\Delta}{\tau}\right)} e^{jw\Delta} + e^{2jw\Delta} - 2e^{-\left(\frac{\Delta}{\tau}\right)} e^{-jw\Delta} + 4 - 2e^{-\left(\frac{-\Delta}{\tau}\right)} e^{jw\Delta} + e^{-2jw\Delta} - 2e^{-\left(\frac{\Delta}{\tau}\right)} e^{-jw\Delta} + e^{-\left(\frac{2\Delta}{\tau}\right)} \right)$$

$$= \frac{w^2 \tau^2}{(1+w^2 \tau^2)^2} \left(\frac{\tau}{e\Delta}\right)^2 \left( 4 + e^{\left(\frac{2\Delta}{\tau}\right)} + e^{-\left(\frac{2\Delta}{\tau}\right)} - 2 \left( e^{\left(\frac{\Delta}{\tau}\right)} + e^{-\left(\frac{\Delta}{\tau}\right)} \right) \left( e^{jw\Delta} + e^{-jw\Delta} \right) + e^{2jw\Delta} + e^{-2jw\Delta} \right)$$

$$= \frac{w^2 \tau^2}{(1+w^2 \tau^2)^2} \left(\frac{\tau}{e\Delta}\right)^2 \left( \left( e^{\left(\frac{\Delta}{\tau}\right)} + e^{-\left(\frac{\Delta}{\tau}\right)} \right)^2 - 4 \left( e^{\left(\frac{\Delta}{\tau}\right)} + e^{-\left(\frac{\Delta}{\tau}\right)} \right) (\cos w\Delta) + 4(\cos w\Delta)^2 \right)$$

$$= \frac{w^2 \tau^2}{(1+w^2 \tau^2)^2} \left(\frac{\tau}{e\Delta}\right)^2 \left( e^{\left(\frac{\Delta}{\tau}\right)} + e^{-\left(\frac{\Delta}{\tau}\right)} - 2\cos w\Delta \right)^2$$

$$\sigma_{sum}^2(t_\infty) = \int_0^\infty \frac{Q_f}{w} \left(\frac{A c_{tot}}{gm(c_{tot} + A c_f)}\right)^2 \frac{w^2 \tau^2}{(1+w^2 \tau^2)^2} \left(\frac{\tau}{e\Delta}\right)^2 \left( e^{\left(\frac{\Delta}{\tau}\right)} + e^{-\left(\frac{\Delta}{\tau}\right)} - 2\cos w\Delta \right)^2 dw$$

substituting  $x = w\tau$

$$\sigma_{sum}^2(t_\infty) = Q_f \frac{1}{e^2} \left(\frac{A c_{tot}}{gm(c_{tot} + A c_f)}\right)^2 \int_0^\infty \frac{x}{(1+x^2)^2} \left(\frac{\tau}{\Delta}\right)^2 \left( e^{\left(\frac{\Delta}{\tau}\right)} + e^{-\left(\frac{\Delta}{\tau}\right)} - 2\cos x \frac{\Delta}{\tau} \right)^2 dx$$

If  $\tau = 2\Delta$

$$\sigma_{sum}^2(t_\infty) = Q_f \left(\frac{A c_{tot}}{gm(c_{tot} + A c_f)}\right)^2 \frac{2.95}{e^2} \text{ for FET input } \sigma_{sum}^2(t_\infty) = \frac{Kf_2}{WLC_{ox}} \left(\frac{A c_{tot}}{(c_{tot} + A c_f)}\right)^2 \frac{2.95}{e^2}$$